

SERVICE MANUAL M113

(SUPER ONE CHIP)

TCL TECHNOLOGY

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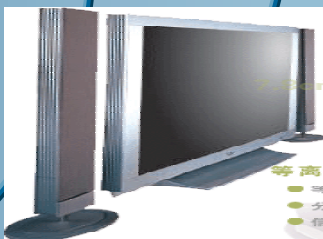


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TCL

SERVICE MANUAL M113 (SUPER ONE-CHIP CHASSIS)

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SERVICE MANUAL FOR M113 CHASSIS

PART I. Servicing Precautions

When working, the unit is with ultra high voltage about **25KV** inside. So, to avoid the risk of electric shock, be careful to adjust the chassis!

1. Only qualified personnel should perform service procedures.
2. All specification must be met over line voltage ranger of 110V AC to 240V AC 50Hz/60Hz.
3. Do not operate in WET/DAMP conditions.
4. Portions of the power supply board are hot ground. The remaining boards are cold ground.
5. Discharge of CRT anode should be done only to CRT ground strap.
6. When fuse blow, ensure to replace a fuse with the same type and specification.
7. Keep the wires away from the components with high temperature or high voltage.
8. When replacing the resister with high power, keep it over the PCB about 10mm.
9. The CRT anode high voltage has been adjusted and set in the factory. When repairing the chassis, do not make the high voltage exceed 27.5KV (The beam current is 0uA). Generally, the high voltage is set on $25.5KV \pm 1.5KV$ (The beam current is 700uA).
* The values of parameters above are for information only.
10. Before return the fixed unit, do check all the covering of wires to ensure that not fold or not short with any metal components. Check the entire protection units, such as control knobs, rear cabinet & front panel, insulation resister & capacitor, mechanical insulators and so on.
11. There are some mechanical and electrical parts associating with safety (EMC) features (Generally related to high voltage or high temperature or electric shock), these features cannot be found out from the outside. When replace these components, perhaps the voltage and power suit the requirements, but efficient X-ray protection may not be provided. All these components are marked with Δ in the schematic diagram. When replace these, you'd better look up the components listed in this manual. If the component you replaced not has the same safety (EMC) performance, harmful X-ray may be produced.

Part II -Product Specification**1. Ambient conditions:**

1.1 Ambient temperatures:

a. Operating: -10°C ~ +40°C

b. Storage: -15°C ~ +45°C

1.2 Humidity

a. Operation: < 80%

b. Storage: < 90%

1.3 Air pressure: 86kpa ~ 106kpa

2. M113 Chassis Specification

2.1 MCU&Chroma Decoder:TMPA8809 Super one chip IC

2.2 System

PAL DK/BG/I

SECAM DK/BG

NTSC 3.579/4.43 AV MODE

Receiving channels 48.25MHz — 463.25MHz (Hyper band)

471.25MHz — 855.25MHz (UHF)

Scanning lines and frequencies 525/625 lines 15.625kHz/15.75kHz 50/60Hz

Color sub-carrier 4.433MHz/3.579MHz

2.3 IF:picture 38.0MHz sound 5.5MHz/6.0MHz/6.5MHz

2.4 Power Consumption:80W

2.5 Power Supply:AC 220V 45-55Hz

2.6 Audio Output Power(7%THD):6W+6W

2.7 Aerial Input Impedance:75Ω Unbalanced Din Jack Ant.Input

2.8 Product EMC/EMI Requirement:CA

2.9 Product EMC/EMI Requirement:CA

3. Basic Feature of Controller

3.1 Channel Tuning Method:Voltage Synthesizer

3.2 Presettable Program:100 Programs

3.3 Tuning for VHF and UHF Bands:Auto/Manual/Fine Tuning

3.4 Picture and Sound Adjustment

Bright, Contrast, Color and Volume Control
 Tint Control(NTSC)
 Treble, Bass, Balance Control
 Sharpness Control

3.5 OSD

General Features(Volume, Brightness, Contrast, Color, Program, Band, AutoSearch, Manual, Tune, Muting, AV And Sleep Timer)
 Stereo Dual Language
 Four Sound Effect Indicator
 German Stereo Indicator

3.6 Sleep Timer:15MIN

3.7 Remote Effective Distance:8m

3.8 Construction of Front Panel

Main Power Switch
 Remote Sensor
 Menu Select
 S.VHS Input
 TV/AV Select
 Standby Indicator
 Program Volume UP/DOWN
 RCA Socket

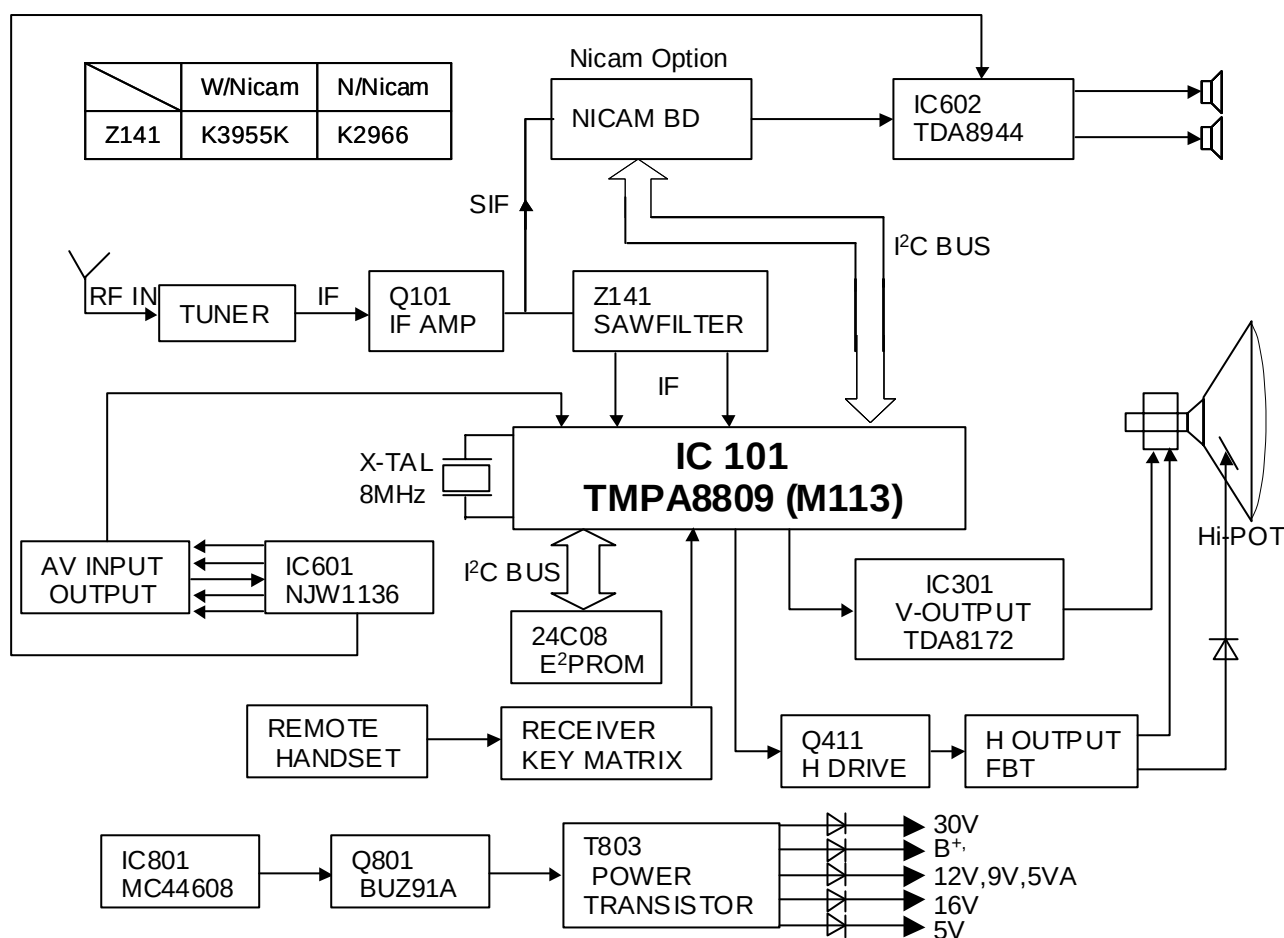
3.9 Construction of Rear Panel

75 Ω Aerial Terminal
 RCA Socket-Audio-R+L In/Out, Video-In/Out
 Y/U/V input

Specification	Scart	RCA
Video input 75 Ω	1Vp-p	1Vp-p
Audio input 10k Ω (R+L)	0.5Vrms	0.5Vrms
Video output 75 Ω	1Vp-p	1Vp-p
Audio output 1k Ω (R+L)	0.5Vrms	0.5Vrms
RGB input 75 Ω	0.7Vp-p	
Audio line output 1k Ω	1Vp-p	

- Design and specifications are subject to change without prior notice for the purpose of performance improvement.
- This specification is only for your reference.

Part III-Brief Introduction on Chassis

**1.RF、 IF AMP**

The tuner receives, selects and amplifies RF signal, frequency mixes with local oscillate source, gets 38.0MHz&38.9MHz IF signal via C107 to Q101. After Q101 amplifying about 20dB, the PIF(Picture IF) and SIF(Sound IF) are separated. Having passed sawfilter, PIF signal sent to TMPA8809 in pin 41,42. The IF signal pass the video detect circuit to generate CVBS signal. Then the processor deals the signal with luminance and chroma separation. The processor also deals the chroma signal with integrated chroma BPFs, PAL/NTSC demodulation and deals luminance signal with integrated chroma traps, black stretch, Y-gamma, so that the resolution of picture details is improved and Y signal is well timed with chroma signal. The processor also deals the chroma signal with chroma sub-carrier recovery, color system recognition and color signal decoding, then output R\G\B to CRT board. Via three groups dual emitter amplifiers to drive KR\KG\KB. On the other hand, the processor separated by video detect circuit. Having passed the horizontal & vertical frequency dividing circuit, H&V OSC signal, which be generated by H-AFC&V-AFC, then output H&V signal which wave is sawtooth.

2.Channel Selection

The RF signal is converted into IF signal by the tuner. Then the IF signal cross the IF amplifier circuit(IF pre-amp) to get a gain about 15dB. By the coupling capacitance(C107) and the match resistance R107, the input resistance of the IF pre-amplifier match with the tuner. The signals pass a parallel connection circuit with voltage NFB, which the input and output impedance is lower, of wide dynamic range. R106 is the NFB resistance, which is used to adjust the gain in the pass band. Having been amplified by the IF amplifier, the IF signal pass the IF sawfilter K3955K(and C109 is the coupling cap.). Then PIF signal been sent from pin 4, pin5 of sawfilter to pin 41, pin42 of super one chip(TMPA8809). The processor deal the PIF signal with IF

detection, PLL demodulation, IF AGC, AFC, video peak detection, and color system recognition ect, then output a AGC signal from pin43 to the tuner to adjust the input control IF detection. R228, C227, C226 makes up picture IF PLL circuit, which is used to control IF detection. IC101 output a TV signal from pin30, when pin30 level is high, TV-out signal is amplified by Q202, Q204, and Q203, Q205 is system switch which controlled by Q211, Q210.

Tuning control and band switch control circuits

The processor output a tuning control signal from pin60. The control signal will pass Q004 and R/C network to be amplified and differential circuit, then added to VT terminal to provide all channels' tuning voltage for the tuner to make the channel stable.

3. Vertical Output Section

TMPA8809 outputs vertical saw-tooth wave from pin 16. It come to pin1 of TDA8172 with DC coupling, and is amplified by inner difference amplifier. Pin7 of TDA8172 is the same phase input terminal. R302 and R303 are DC offset resistances. C304 is a filter capacitor. In application to M113, pin7 of TDA8172 is fixed as the DC amplify ref terminal. The amplified saw tooth-wave comes out from pin5 and make the deflect coil to generate the deflect current. R304 and C305 filtrate the inductive interference from the horizontal deflect coil. R310 and C309 are used to eliminate spurious oscillation generated by the deflect coil and distributed capacitance resonance. C308, R309, C307 and accessory circuit are in charge of draw AC saw tooth wave out at the deflect coil terminal connected with R303A, and feedback to the input terminal of TDA8172 (pin1) to correct the linearity of horizontal scan. C301 is a high frequency decoupling capacitor. D301 and C302 make up of a voltage pump up circuit. TDA8172 output a vertical kickback impulse from pin6 to locate the OSD characters.

4. Horizontal Output And FBT Section

The processor outputs horizontal drive impulse from pin 13(H-OUT). The drive impulse is done with voltage division by R201 and R401, and then comes to the base of the drive triode (Q401). C401 is used to eliminate the noise in the H drive impulse. T401 is a horizontal drive transformer. Q411 is a horizontal output triode with a damper inside. L412 is connected with the emitter of the horizontal output diode to eliminate the radiation and to improve the distortions at the cross of vertical and horizontal white lines. C412 and C413 are retrace capacitors and C414 is an s-correct capacitor. L413 is horizontal linear inductors. R441 is used to eliminate the parasitic oscillation caused by horizontal linear inductors. C421, R421 and D421 are used to correct the M-distortion in horizontal direction.

The deflect coil and the horizontal output triode have some resistance R while they are ducting. The resistance R will cause the non-linear distortion, which means that the right direction scanning speed of the electron beam becomes slower, and the right of the raster is compressed to generate distortion. We use a horizontal linear adjuster to compensate this kind of distortion. We use L414 as the H linear adjuster in horizontal scanning section of M113 chassis. R419, which is parallel connected with L414, is a despiing resistance for preventing the oscillation by compensating inductor and the stray capacitance. The linear adjuster is a transducer coil with a magnetic core inside. If the current, which pass the linear adjuster coil, increase to a certain value, the magnetic core becomes saturated to decrease the inductance of the linear adjustment inductor. If the +B is steady, the increase speed of Iy is faster to compensate the reducing of deflecting current by the resistance R mention above.

We can adjust the magnetic core to change the inductance of the linear compensate inductor to adjust the H linearity.

The EW-correct signal sent from pin28 of TMPA8809, amplified with Q412, Q413 and Q414, to adjust

horizontal output circuit.

The EHT generation circuit

The FBT supply the anode high voltage, focus voltage and screen voltage for M113 chassis. D441 and C441 are in charge of regulating the primary impulse of the transformer to output a voltage of 200V for the video amplifiers. The (4) ~ (7) coils of the FBT supply the heater with power. To limit the beam current in a safe range, we add a ABL(auto brightness limit) circuit in M113 chassis. When the beam current is higher than normal, Q451 which is an emitter follower strength conductivity, the emitter gets a lower negative voltage, so the collector of it follows a lower voltage, then gain of system brightness decreases, brightness decrease and beam current decreases.

Also the ABL control voltage is sampled from R426 to adjust & control EW-scan.

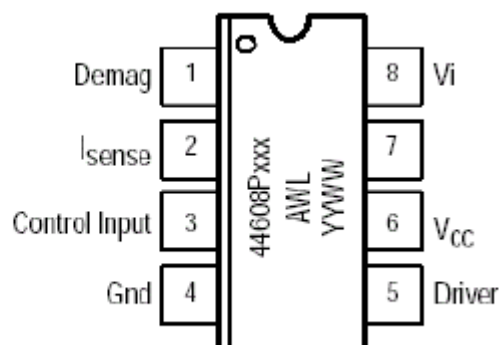
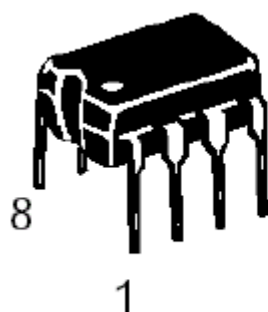
Extension distortion and compensation

This kind of distortion is mainly caused by the structure of CRT. Due to the screen of SF CRT is not a pure flat screen, the distances from the deflecting center to the screen are not the same. The scanning speed of the electron beam is uniform. If the electron beam scanning the screen equally with the effect of a true linear sawtooth current, the E-W sides of the picture are stretched. That is the extension distortion. Usually, we add a S-correct capacitor in series with the deflecting coil to compensate this kind of distortion. The integral character of S-correct capacitor makes the current waveform S shape. So the scanning speed of electron beam at the center of screen is faster than the one at the side. So this action can correct the extension distortion. C414 is a S-correct capacitor. The capacitance is inverse ratio with the correcting effect.

PART IV. IC Pin Description

1. MC44608-High Voltage PWM Controller

Pin	Name	Description description
1	Demag	The Demag pin offers 3 different functions: Zero voltage crossing detection (50mV), 24 A current detection and 120 A current detection. The 24 A level is used to detect the secondary reconfiguration status and the 120 A level to detect an Over Voltage status called Quick OVP.
2	I _{sense}	The Current Sense pin senses the voltage developed on the series resistor inserted in the source of the power MOSFET. When I _{sense} reaches 1V, the Driver output (pin 5) is disabled. This is known as the Over Current Protection function. A 200 A current source is flowing out of the pin 3 during the start-up phase and during the switching phase in case of the Pulsed Mode of operation. A resistor can be inserted between the sense resistor and the pin 3, thus a programmable peak current detection can be performed during the SMPS stand-by mode.
3	Control Input	A feedback current from the secondary side of the SMPS via the opto-coupler is injected into this pin. A resistor can be connected between this pin and GND to allow the programming of the Burst duty cycle during the Stand-by mode.
4	Ground	This pin is the ground of the primary side of the SMPS.
5	Driver	The current and slew rate capability of this pin are suited to drive Power MOSFETs.
6	V _{CC}	This pin is the positive supply of the IC. The driver output gets disabled when the voltage becomes higher than 15V and the operating range is between 6.6V and 13V. An intermediate voltage level of 10V creates a disabling condition called Latched Off phase.
7		This pin is to provide isolation between the V _i pin 8 and the VCC pin 6.
8	V _i	This pin can be directly connected to a 500V voltage source for start-up function of the IC. During the Start-up phase a 9 mA current source is internally delivered to the VCC pin 6 allowing a rapid charge of the VCC capacitor. As soon as the IC starts-up, this current source is disabled.



OPERATING DESCRIPTION

Regulation

The pin 3 senses the feedback current provided by the opto-coupler. During the switching phase the switch S2 is closed and the shunt regulator is accessible by the pin 3. The shunt regulator voltage is typically 5V. The dynamic resistance of the shunt regulator represented by the zener diode is 20 .

The gain of the Control input is given on Figure 10 which shows the duty cycle as a function of the current injected into the pin 3.

The maximum current sense threshold is fixed at 1V. The peak

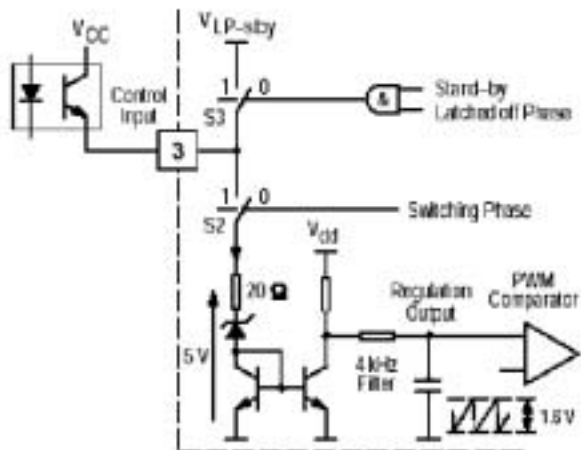


Figure 1. Regulator

A 4KHz filter network is inserted between the shunt regulator and the PWM comparator to cancel the high frequency residual noise.

The switch S3 is closed in Stand-by mode during the Latched Off Phase while the switch S2 remains open. (See section PULSED MODE DUTY CYCLE CONTROL).

The resistor Rdpulsed (Rduty cycle burst) has no effect on the regulation process. This resistor is used to

determine the burst duty cycle described in the chapter “Pulsed Duty Cycle Control” on page 8.

PWM Latch

The MC44608 works in voltage mode. The on-time is controlled by the PWM comparator that compares the oscillator sawtooth with the regulation block output.

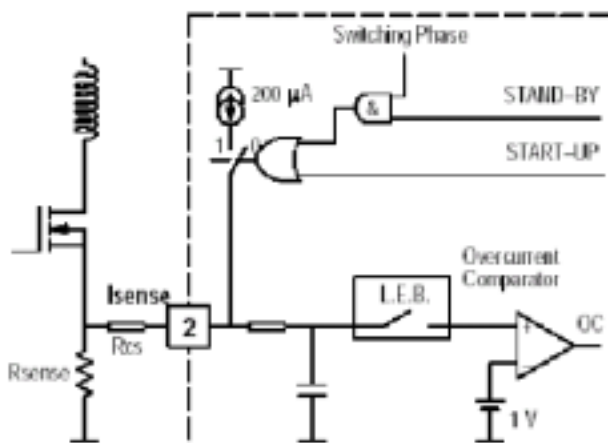


Figure 2. Current Sense

The PWM latch is initialized by the oscillator and is reset by the PWM comparator or by the current sense comparator in case of an over current. This configuration ensures that only a single pulse appears at the circuit output during an oscillator cycle.

Current Sense

The inductor current is converted to a positive voltage by inserting a ground reference sense resistor R_{Sense} in series with the power switch.

The maximum current sense threshold is fixed at 1V. The peak current is given by the following equation:

$$I_{pkmax} = 1/R_{sense} \text{ (A)}$$

In stand-by mode, this current can be lowered as due to the activation of a 200 A current source:

$$I_{pkMAX-STBY}$$

The current sense input consists of a filter (6k , 4pF) and of a leading edge blanking. Thanks to that, this pin is not sensitive to the power switch turn on noise and spikes and practically in most applications, no filtering network is

required to sense the current.

Finally, this pin is used:

- as a protection against over currents ($I_{\text{sense}} > I$)
- as a reduction of the peak current during a Pulsed Mode switching phase.

The overcurrent propagation delay is reduced by producing a sharp output turn off (high slew rate).

This results in an abrupt output turn off in the event of an over current and in the majority of the pulsed mode switching sequence.

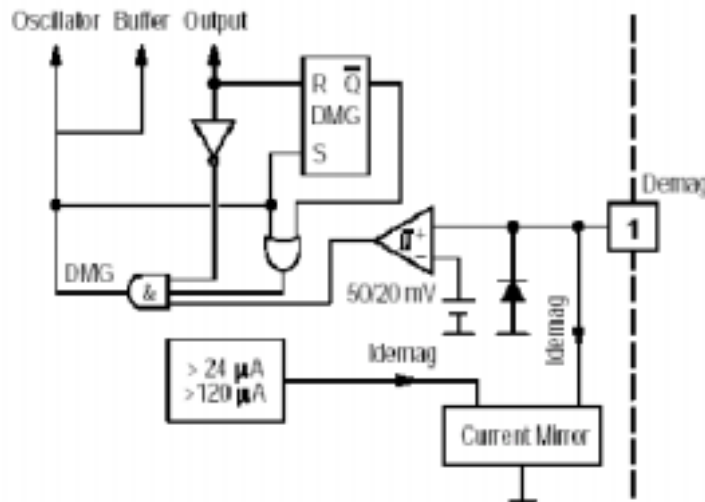


Figure 3. Demagnetization Block

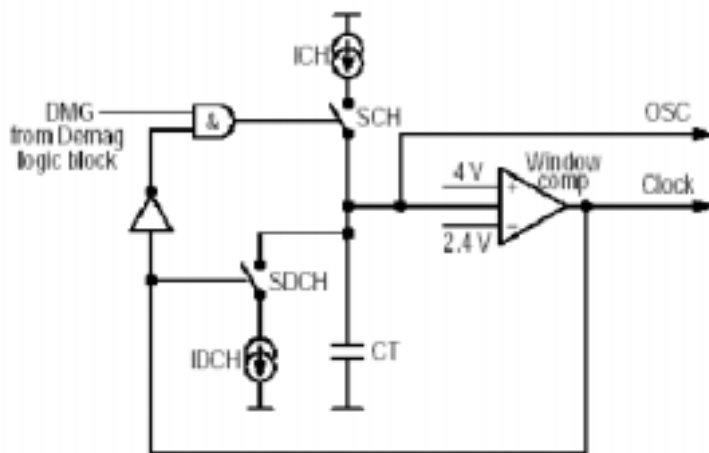


Figure 4. Oscillator Block

Demagnetization Section

The MC44608 demagnetization detection consists of a comparator designed to compare the V_{CC} winding voltage to a reference that is typically equal to 50mV.

This reference is chosen low to increase effectiveness of the demagnetization detection even during start-up.

A latch is incorporated to turn the demagnetization block output into a low level as soon as a voltage less than 50 mV is detected, and to keep it in this state until a new pulse is generated on the output. This avoids any ringing on the input signal which may alter the demagnetization detection.

For a higher safety, the demagnetization block output is also directly connected to the

output, which is disabled during the demagnetization phase.

The demagnetization pin is also used for the quick, programmable OVP. In fact, the demagnetization input current is sensed so that the circuit output is latched off when this current is detected as higher than 120 μ A.

This function can be inhibited by grounding it but in this case, the quick and programmable OVP is also disabled.

Oscillator

The MC44608 contains a fixed frequency oscillator. It is built around a fixed value capacitor CT successively charged and discharged by two distinct current sources ICH and IDCH. The window comparator senses the CT voltage value and activates the sources when the voltage is reaching the 2.4V/4V levels.

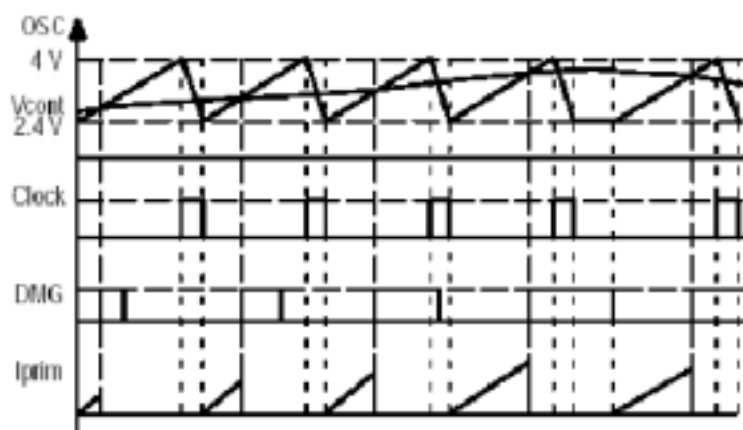


Figure 5.

Figure 5).

The OSC and Clock signals are provided according to the Figure 5. The Clock signals correspond to the CT capacitor discharge. The bottom curve represents the current flowing in the sense resistor R_{cs} . It starts from zero and stops when the sawtooth value is equal to the control voltage V_{cont} . In this way the SMPS is regulated with a voltage mode control.

Overvoltage Protection

The MC44608 offers two OVP functions:

- a fixed function that detects when V_{cc} is higher than 15.4V
- a programmable function that uses the demag pin. The current flowing into the demag pin is mirrored and compared to the reference current I_{ovp} (120 μ A). Thus this OVP is quicker as it is not impacted by the V_{cc} inertia and is called QOVP.

In both cases, once an OVP condition is detected, the output is latched off until a new circuit

START-UP.

Start-up Management

The Vi pin 8 is directly connected to the HV DC rail V_{in} . This high voltage current source is

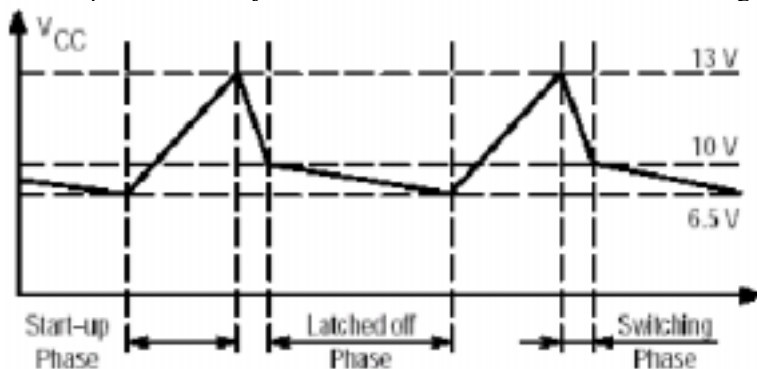


Figure 6. Hiccup Mode

internally connected to the V_{cc} pin and thus is used to charge the V_{cc} capacitor. The V_{cc} capacitor charge period corresponds to the Start-up phase. When the V_{cc} voltage reaches 13V, the high voltage 9mA current source is disabled and the device starts working. The device enters into the switching phase.

It is to be noticed that the maximum rating of the Vi pin 8 is 700V. ESD protection circuitry is not currently added to this pin due to size limitations and technology constraints. Protection is limited by the drain-substrate junction in

avalanche breakdown. To help increase the application safety against high voltage spike on that pin it is possible to insert a small wattage 1k series resistor between the Vin rail and pin 8.

The Figure 6 shows the Vcc voltage evolution in case of no external current source providing current into the Vcc pin during the switching phase. This case can be encountered in SMPS when the self supply through an auxiliary winding is not present (strong overload on the SMPS output for example).

The Figure 16 also depicts this working configuration.

In case of the hiccup mode, the duty cycle of the switching phase is in the range of 10%.

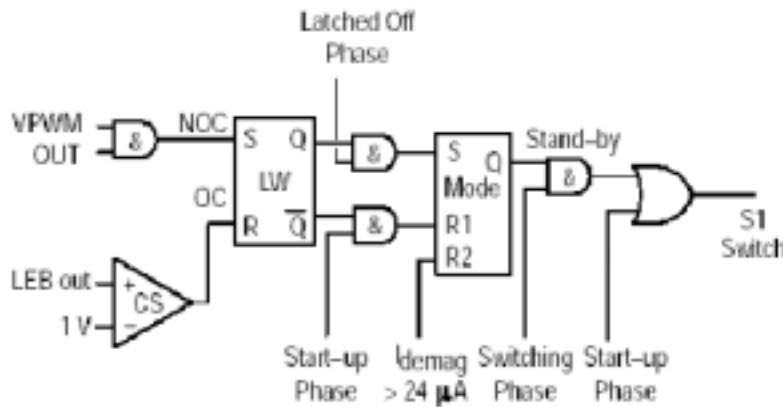


Figure 7. Transition Logic

Mode Transition

The LW latch Figure 7 is the memory of the working status at the end of every switching sequence. Two different cases must be considered for the logic at the termination of the SWITCHING PHASE:

1. No Over Current was observed
2. An Over Current was observed

These 2 cases are corresponding to the signal labeled NOC in case of “No Over Current” and “OC” in case of Over Current. So the effective working status at the end of the ON time memorized in LW corresponds to Q=1 for no over current and Q=0 for over current.

This sequence is repeated during the Switching phase.

Several events can occur:

1. SMPS switch OFF
2. SMPS output overload
3. Transition from Normal to Pulsed Mode
4. Transition from Pulsed Mode to Normal Mode

1. SMPS SWITCH OFF

When the mains is switched OFF, so long as the bulk electrolytic bulk capacitor provides energy to the SMPS, the controller remains in the switching phase. Then the peak current reaches its maximum peak value, the switching frequency decreases and all the secondary voltages are reduced. The Vcc voltage is also reduced. When Vcc is equal to 10V, the SMPS stops working.

2. Overload

In the hiccup mode the 3 distinct phases are described as follows (refer to Figure 6):

The SWITCHING PHASE: The SMPS output is low and the regulation block reacts by increasing the ON time ($d_{max} = 80\%$). The OC is reached at the end of every switching cycle. The LW latch (Figure 7) is reset before the VPWM signal appears. The SMPS output voltage is low. The Vcc voltage cannot be maintained at a normal level as

the auxiliary winding provides a voltage which is also reduced in a ratio similar to the one on the output (i.e. V_{out} nominal / V_{out} short-circuit). Consequently the V_{CC} voltage is reduced at an operating rate given by the combination V_{CC} capacitor value together with the I_{CC} working consumption (3.2mA) according to the equation 2. When V_{CC} crosses 10V the WORKING PHASE gets terminated. The LW latch remains in the reset status.

The LATCHED-OFF PHASE: The V_{CC} capacitor voltage continues to drop. When it reaches 6.5V this phase is terminated. Its duration is governed by equation 3.

The START-UP PHASE is reinitiated. The high voltage start-up current source ($-I_{CC1} = 9mA$) is activated and the MODE latch is reset. The V_{CC} voltage ramps up according to the equation 1. When it reaches 13V, the IC enters into the SWITCHING PHASE.

The NEXT SWITCHING PHASE: The high voltage current source is inhibited, the MODE latch ($Q=0$) activates the NORMAL mode of operation. Figure 2 shows that no current is injected out pin 2.

The over current sense level corresponds to 1V.

As long as the overload is present, this sequence repeats. The SWITCHING PHASE duty cycle is in the range of 10%.

3. Transition from Normal to Pulsed Mode

In this sequence the secondary side is reconfigured (refer to the typical application schematic on page 13). The high voltage output value becomes lower than the NORMAL mode regulated value. The TL431 shunt regulator is fully OFF. In the SMPS stand-by mode all the SMPS outputs are lowered except for the low voltage output that supply the wake-up circuit located at the isolated side of the power supply. In that mode the secondary regulation is performed by the zener diode connected in parallel to the TL431.

The secondary reconfiguration status can be detected on the SMPS primary side by measuring the voltage level present on the auxiliary winding $Laux$. (Refer to the Demagnetization Section). In the reconfigured status, the $Laux$ voltage is also reduced. The V_{CC} self-powering is no longer possible thus the SMPS enters in a hiccup mode similar to the one described under the Overload condition.

In the SMPS stand-by mode the 3 distinct phases are:

The SWITCHING PHASE: Similar to the Overload mode. The current sense clamping level is reduced according to the equation of the current sense section, page 5. The C.S. clamping level depends on the power to be delivered to the load during the SMPS stand-by mode. Every switching sequence ON/OFF is terminated by an OC as long as the secondary Zener diode voltage has not been reached. When the Zener voltage is reached the ON cycle is terminated by a true PWM action. The proper SWITCHING PHASE termination must correspond to a NOC condition. The LW latch stores this NOC status.

The LATCHED OFF PHASE: The MODE latch is set.

The START-UP PHASE is similar to the Overload Mode. The MODE latch remains in its set status ($Q=1$).

The SWITCHING PHASE: The Stand-by signal is validated and the 200 μA is sourced out of the Current Sense pin 2.

4. Transition from Stand-by to Normal

The secondary reconfiguration is removed. The regulation on the low voltage secondary rail can no longer be achieved, thus at the end of the SWITCHING PHASE, no PWM condition can be encountered. The LW latch is reset. At the next WORKING PHASE a NORMAL mode status takes place.

In order to become independent of the recovery time SWITCHING PHASE constant on the secondary side of the

SMPS an additional reset input R2 is provided on the MODE latch. The condition $I_{\text{demag}} < 24 \mu\text{A}$ corresponds to the activation of the secondary reconfiguration status. The R2 reset insures a return into the NORMAL mode following the first corresponds to 1V. START-UP PHASE.

Pulsed Mode Duty Cycle Control

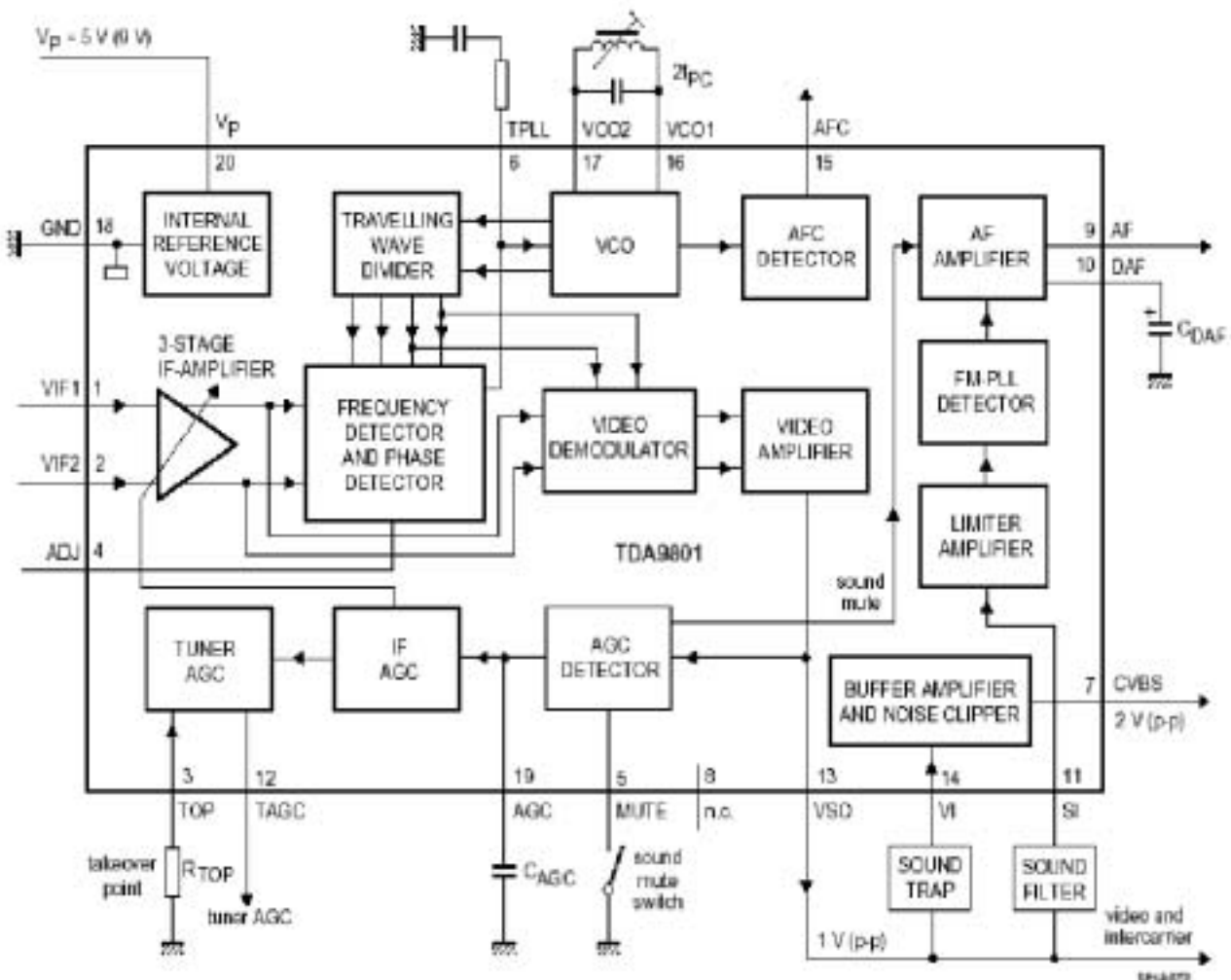
During the sleep mode of the SMPS the switch S3 is closed and the control input pin 3 is connected to a 4.6V voltage source thru a 500 Ω resistor. The discharge rate of the V_{CC} capacitor is given by $I_{\text{CC-latch}}$ (device consumption during the LATCHED OFF phase) in addition to the current drawn out of the pin 3. Connecting a resistor between the Pin 3 and GND (R_{DPULSED}) a programmable current is drawn from the V_{CC} through pin 3. The duration of the LATCHED OFF phase is impacted by the presence of the resistor R_{DPULSED} . The equation 3 shows the relation to the pin 3 current.

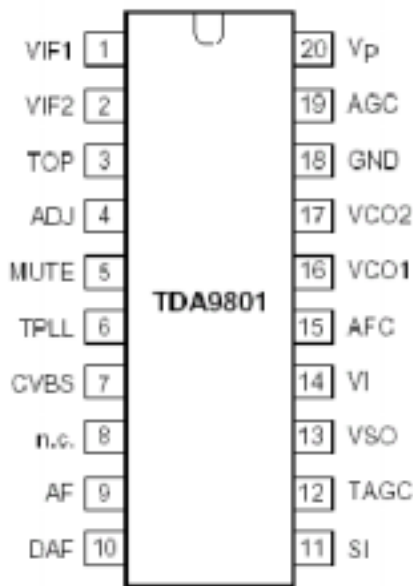
Pulsed Mode Phases

Equations 1 through 8 define and predict the effective behavior during the PULSED MODE operation. The equations 6, 7, and 8 contain K, Y, and D factors. These factors are combinations of measured parameters. They appear in the parameter section “K factors for pulsed mode operation” page 4. In equations 3 through 8 the pin 3 current is the current defined in the above section “Pulsed Mode Duty Cycle Control”.

2. TDA9801-Single standard VIF-PLL demodulator and FM-PLL detector

FUNCTIONAL DESCRIPTION





SYMBOL	PIN	DESCRIPTION
VIF1	1	VIF differential input 1
VIF2	2	VIF differential input 2
TOP	3	tuner AGC TakeOver Point (TOP) connection
ADJ	4	phase adjust connection
MUTE	5	sound mute switch connection
TPLL	6	PLL time constant connection
CVBS	7	CVBS (positive) video output
n.c.	8	not connected
AF	9	AF output
DAF	10	AF amplifier decoupling capacitor connection
SI	11	sound intercarrier input
TAGC	12	tuner AGC output
VSO	13	video and sound intercarrier output
VI	14	buffer amplifier video input
AFC	15	AFC output
VCO1	16	VCO1 reference circuit for 2fPC
VCO2	17	VCO2 reference circuit for 2fPC
GND	18	ground supply (0 V)
AGC	19	AGC detector capacitor connection
VP	20	supply voltage (+5 V)

Stage IF amplifier

The VIF amplifier consists of three AC-coupled differential amplifier stages. Each differential stage comprises a feedback network controlled by emitter degeneration.

AGC detector, IF AGC and tuner AGC

The automatic control voltage to maintain the video output signal at a constant level is generated in accordance with the transmission standard. Since the TDA9801 is suitable for negative modulation only the peak sync pulse level is detected.

The AGC detector charges and discharges capacitor C_{AGC} to set the IF amplifier and tuner gain. The voltage on capacitor C_{AGC} is transferred to an internal IF control signal, and is fed to the tuner AGC to generate the tuner AGC output current on pin TAGC (open-collector output). The tuner AGC takeover point level is set at pin TOP. This allows the tuner to be matched to the SAW filter in order to achieve the optimum IF input level.

Frequency detector and phase detector

The VIF amplifier output signal is fed into a frequency detector and into a phase detector. During acquisition the frequency detector produces a DC current proportional to the frequency difference between the input and the VCO signal. After frequency lock-in the phase detector produces a DC current proportional to the phase difference between the VCO and the input signal. The DC current of either frequency detector or phase detector is converted into a DC voltage via the loop filter which controls the VCO frequency.

Video demodulator

The true synchronous video demodulator is realized by a linear multiplier which is designed for low distortion and wide bandwidth. The vision IF input signal is multiplied with the 'in phase' component of the VCO output. The demodulator output signal is fed via an integrated low-pass filter ($f_g = 12$ MHz) for suppression of the carrier harmonics to the video amplifier.

VCO, AFC detector and travelling wave divider

The VCO operates with a symmetrically connected reference LC circuit, operating at the double vision carrier frequency. Frequency control is performed by an internal variable capacitor diode.

The voltage to set the VCO frequency to the actual double vision carrier frequency is also amplified and converted for the AFC output current.

The VCO signal is divided-by-2 with a Travelling Wave Divider (TWD) which generates two differential output signals with a 90 degree phase difference independent of the frequency.

Video amplifier

The composite video amplifier is a wide bandwidth operational amplifier with internal feedback. A nominal positive video signal of 1 V (p-p) is present at pin VSO.

Buffer amplifier and noise clipper

The input impedance of the 7 dB wideband CVBS buffer amplifier (with internal feedback) is suitable for ceramic sound trap filters. Pin CVBS provides a positive video signal of 2 V (p-p). Noise clipping is provided internally.

Sound demodulation

LIMITER AMPLIFIER

The FM sound intercarrier signal is fed to pin SI and through a limiter amplifier before it is demodulated. The result is high sensitivity and AM suppression. The limiter amplifier consists of 7 stages which are internally AC-coupled in order to minimizing the DC offset.

FM-PLL DETECTOR

The FM-PLL demodulator consists of an RC oscillator, loop filter and phase detector. The oscillator frequency is locked on the FM intercarrier signal from the limiter amplifier. As a result of this locking, the RC oscillator is frequency modulated. The modulating voltage (AF signal) is used to control the oscillator frequency. By this, the FM-PLL operates as an FM demodulator.

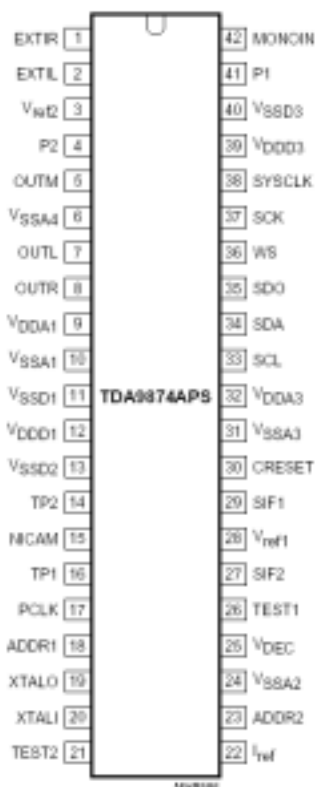
AF AMPLIFIER

The audio frequency amplifier with internal feedback is designed for high gain and high common-mode rejection. The low-level AF signal output from the FM-PLL demodulator is amplified and buffered in a low-ohmic audio output stage. An external decoupling capacitor CDAF removes the DC voltage from the audio amplifier input. By using the sound mute switch (pin MUTE) the AF amplifier is set in the mute state.

3. TDA9874A Digital TV sound demodulator/decoder

SYMBOL	PIN	DESCRIPTION
EXTIR	1	external audio input right channel
EXTIL	2	external audio input left channel
Vref2	3	analog reference voltage for DAC and operational amplifiers
P2	4	second general purpose I/O pin
OUTM	5	analog output mono
VSSA4	6	analog ground supply 4 for analog back-end circuitry
OUTL	7	analog output left
OUTR	8	analog output right
VDDA1	9	analog supply voltage 1; back-end circuitry 5 V
VSSA1	10	analog ground supply 1; back-end circuitry

VSSD1	11	digital ground supply 1; core circuitry
VDDD1	12	digital supply voltage 1; core voltage regulator circuitry
VSSD2	13	digital ground supply 2; core circuitry
TP2	14	additional test pin 2; connected to VSSD for normal operation
NICAM	15	serial NICAM data output (at 728 kHz)
TP1	16	additional test pin 1; connected to VSSD for normal operation
PCLK	17	NICAM clock output (at 728 kHz)
ADDR1	18	first I ² C-bus slave address modifier input
XTALO	19	crystal oscillator output
XTALI	20	crystal oscillator input
TEST2	21	test pin 2; connected to VSSD for normal operation
Iref	22	resistor for reference current generation; front-end circuitry
ADDR2	23	second I ² C-bus slave address modifier input
VSSA2	24	analog ground supply 2; analog front-end circuitry
VDEC	25	analog front-end circuitry supply voltage decoupling
TEST1	26	test pin 1; connected to VSSD for normal operation
SIF2	27	sound IF input 2
Vref1	28	reference voltage; for analog front-end circuitry
SIF1	29	sound IF input 1
CRESET	30	capacitor for Power-on reset
VSSA3	31	digital ground supply 3; front-end circuitry
VDDA3	32	analog front-end circuitry regulator supply voltage 3 (5 V)
SCL	33	I ² C-bus serial clock input
SDA	34	I ² C-bus serial data input/output
SDO	35	I ² S-bus serial data output
WS	36	I ² S-bus word select input/output
SCK	37	I ² S-bus clock input/output
SYSClk	38	system clock output
VDDD3	39	digital supply voltage 3; digital I/O pads
VSSD3	40	digital ground supply 3; digital I/O pads
P1	41	first general purpose I/O pin
MONOIN	42	analog mono input



FUNCTIONAL DESCRIPTION

Description of the demodulator and decoder section

1. SIF INPUTS

Two inputs are provided, pin SIF1 and pin SIF2. For higher SIF signal levels the SIF input can be attenuated with an internal switchable 10 dB resistor divider.

As no specific filters are integrated, both inputs have the same specification giving flexibility in application. The selected signal is passed through an AGC circuit and then digitized by an 8-bit ADC operating at 24.576 MHz.

2. AGC

The gain of the AGC amplifier is controlled from the ADC output by means of a digital control loop employing hysteresis. The AGC has a fast attack behaviour to prevent ADC overloads, and a slow decay behaviour to prevent AGC oscillations. For AM demodulation the AGC must be switched off. When switched off, the control loop is reset and fixed gain settings can be chosen. The AGC can be controlled via the I²C-bus.

3. MIXER

The digitized input signal is fed to the mixers, which mix one or both input sound

carriers down to zero IF. A 24-bit control word for each carrier sets the required frequency. Access to the mixer control word registers is via the I²C-bus or via Easy Standard Programming (ESP). When receiving NICAM programs, a feedback signal is added to the control word of the second carrier mixer to establish a carrier-frequency loop.

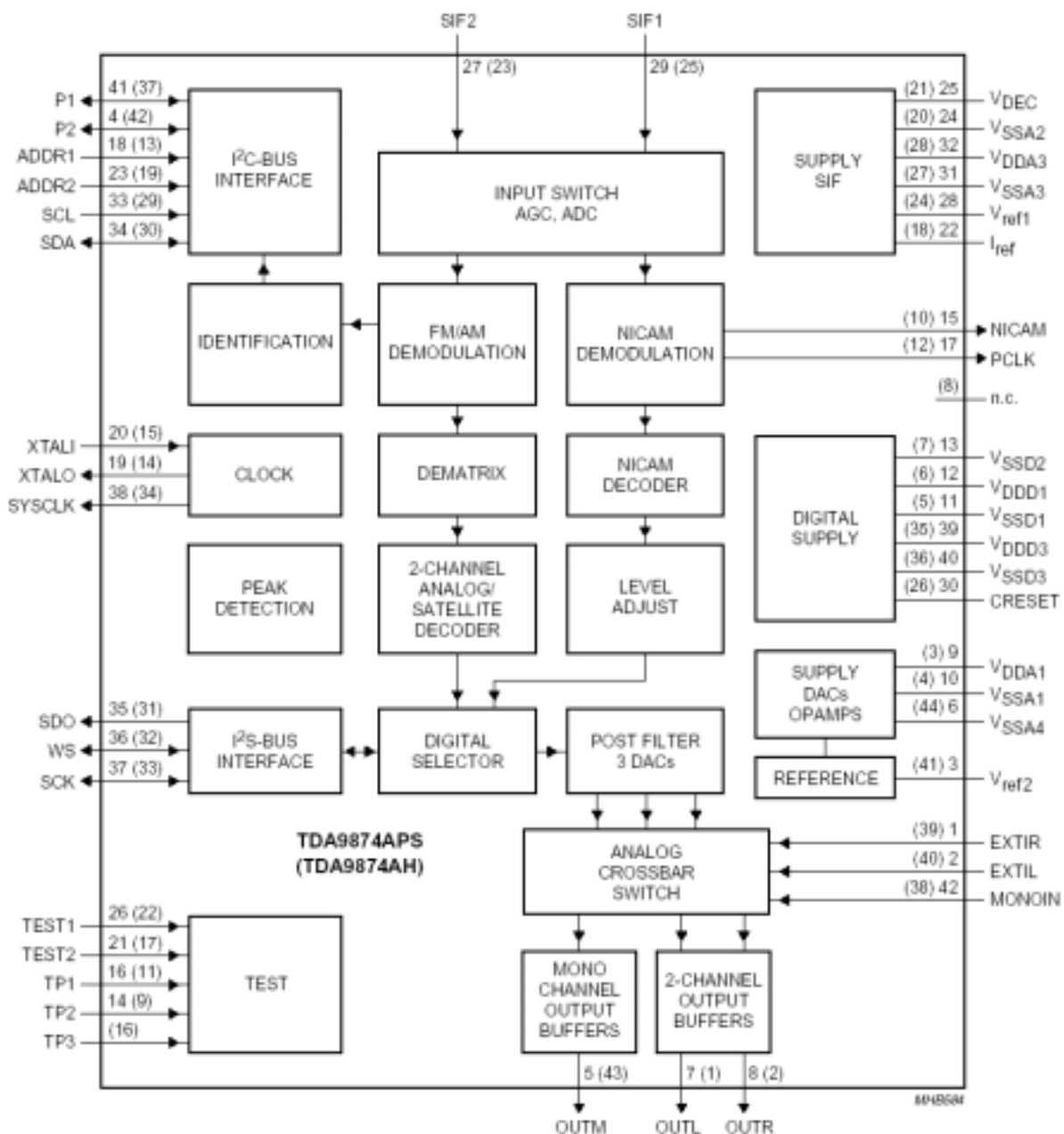
4. FM AND AM DEMODULATION

An FM or AM input signal is fed through a switchable band-limiting filter into a demodulator that can be used for either FM or AM demodulation. Apart from the standard (fixed) de-emphasis characteristic, an adaptive de-emphasis is available for Wegener-Panda 1 encoded satellite programs.

5. FM DECODING

A 2-carrier stereo decoder recovers the left and right signal channels from the demodulated sound carriers. Both the European and Korean stereo systems are supported.

Automatic FM dematrixing is also supported, which means that the FM sound mode identification (mono, stereo or dual) switches the FM dematrix directly. No loop via the microcontroller is needed.



For highly overmodulated signals, a high deviation mode for monaural audio sound single carrier demodulation can be selected.

NICAM decoding is still possible in high deviation mode.

6. FM IDENTIFICATION

The identification of the FM sound mode is performed by AM synchronous demodulation of the pilot and narrow-band detection of the identification frequencies. The result is available via the I2C-bus interface. A selection can be made via the I2C-bus for B/G, D/K and M standards, and for three different time constants that represent different trade-offs between speed and reliability of identification. A pilot detector allows the control software to identify an analog 2-carrier (A2) transmission within approximately 0.1 s.

Automatic FM dematrixing, depending on the identification, is possible.

7. NICAM DEMODULATION

The NICAM signal is transmitted in a DQPSK code at a bit rate of 728 kbits/s. The NICAM demodulator performs DQPSK demodulation and passes the resulting bitstream and clock signal to the NICAM decoder and, for evaluation purposes, to various pins.

A timing loop controls the frequency of the crystal oscillator to lock the sampling instants to the symbol timing of the NICAM data.

8. NICAM DECODING

The device performs all decoding functions in accordance with the “EBU NICAM 728 specification”. After locking to the frame alignment word, the data is descrambled by applying the defined pseudo-random binary sequence. The device then synchronizes to the periodic frame flag bit C0.

The status of the NICAM decoder can be read out from the NICAM status register by the user. The OSB bit indicates that the decoder has locked to the NICAM data. The VDSP bit indicates that the decoder has locked to the NICAM data and that the data is valid sound data. The C4 bit indicates that the sound conveyed by the FM mono channel is identical to the sound conveyed by the NICAM channel.

The error byte contains the number of sound sample errors (resulting from parity checking) that occurred in the past 128 ms period. The Bit Error Rate (BER) can be calculated using the following equation:

$$\text{BER} = \text{bit errors} / \text{total bits} \approx \text{error byte} \times 1.74 \times 10^{-5}$$

9. NICAM AUTO-MUTE

This function is enabled by setting bit AMUTE to logic 0. Upper and lower error limits may be defined by writing appropriate values to two registers in the I2C-bus section. When the number of errors in a 128 ms period exceeds the upper error limit, the auto-mute function will switch the output sound from NICAM to whatever sound is on the first sound carrier (FM or AM) or to the analog mono input. When the error count is smaller than the lower error limit, the NICAM sound is restored.

The auto-mute function can be disabled by setting bit AMUTE to logic 1. In this case clicks become audible when the error count increases. The user will hear a signal of degrading quality.

If no NICAM sound is received, the outputs are switched from the NICAM channel to the 1st sound carrier.

A decision to enable or disable the auto-mute is taken by the microprocessor based on an interpretation of the application control bits C1, C2, C3 and C4, and possibly any additional strategy implemented by the user in the microcontroller software. When the AM sound in NICAM L systems is demodulated in the 1st sound IF and the audio signal connected to the mono input of the TDA9874A, the controlling microprocessor has to ensure switching from NICAM reception to mono input, if auto-muting is desired. This can be achieved by setting bit AMSEL = 1 and bit AMUTE = 0.

10. CRYSTAL OSCILLATOR

The digital controlled crystal oscillator (DCXO) is fully integrated. Only an external 24.576 MHz crystal is required.

11. TEST PINS

All test pins are active HIGH. In normal operation of the device they can be left open-circuit, as they have internal pull-down resistors. Test functions are for manufacturing tests only and are not available to customers.

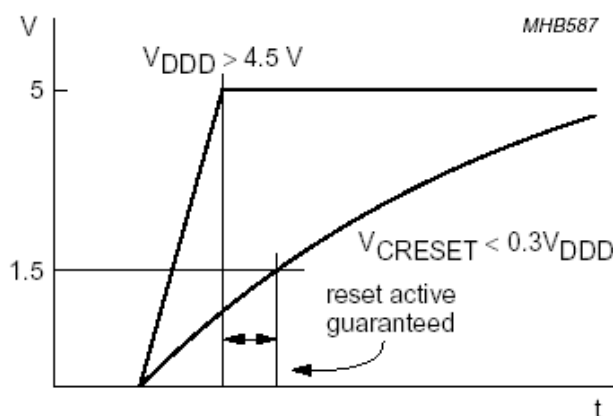
12. POWER FAIL DETECTOR

The power fail detector monitors the internal power supply for the digital part of the device. If the supply has temporarily been lower than the specified lower limit, the power failure register bit PFR in subaddress 0, will be set to logic 1. Bit CLRPFR, slave register subaddress 1, resets the Power-on reset flip-flop to logic 0. If this is detected, an initialization of the TDA9874A has to be performed to ensure reliable operation.

13. POWER-ON RESET

The reset is active LOW. In order to perform a reset at power-up, a simple RC circuit may be used which consists of an integrated passive pull-up resistor and an external capacitor connected to ground.

The pull-up resistor has a nominal value of 50 k, which can easily be measured between pins CRESET and VDDD3. Before the supply voltage has reached a certain minimum level, the state of the circuit is completely undefined and remains in this undefined state until a reset is applied.



The reset is guaranteed to be active when:

- .The power supply is within the specified limits (4.5 to 5.5 V)
- .The crystal oscillator (DCXO) is functioning
- .The voltage at pin CRESET is below 0.3VDDD (1.5 V if VDDD= 5.0 V, typically below 1.8 V).

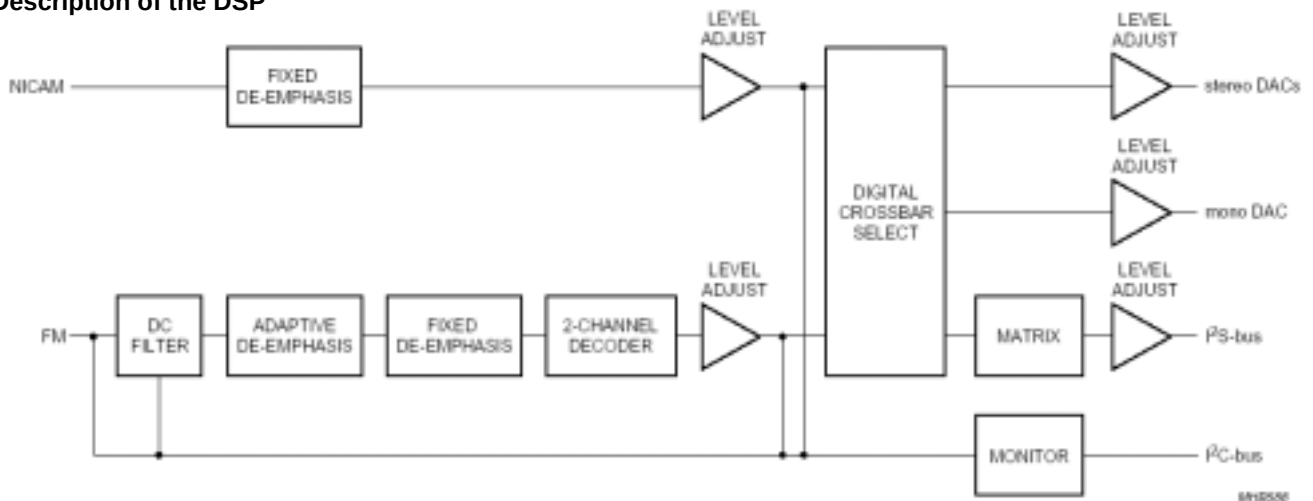
The required capacitor value depends on the gradient of the rising power supply voltage. The time constant of the RC circuit should be clearly larger than the rise time of the power supply (to make sure that the reset condition is

always satisfied), even when considering tolerance spreading. To avoid problems with a too slow discharging of the capacitor at

power-down, it may be helpful to add a diode from pin CRESET to VDDD.

It should be noted that the internal ESD protection diode does not help here as it only conducts at higher voltages. Under difficult power supply conditions (e.g. very slow or non-monotonic ramp-up), it is recommended to drive the reset line from a microcontroller port or the like.

Description of the DSP



1. LEVEL SCALING

All input channels to the digital crossbar switch are equipped with a level adjustment facility to change the signal level in a range of ± 15 dB. Adjusting the signal level is intended to compensate for the different modulation parameters of the various TV standards. Under nominal conditions it is recommended to scale all input channels to be 15 dB below full-scale. This will create sufficient headroom to cope with overmodulation and avoids changes of the volume impression when switching from FM to NICAM or vice versa.

2. NICAM PATH

The NICAM path has a switchable J17 de-emphasis.

3. NICAM AUTO-MUTE

If NICAM is received, the auto-mute is enabled and the signal quality becomes poor. The digital crossbar switches automatically to FM, channel 1 or the analog mono input, as selected by bit AMSEL. This automatic switching depends on the NICAM bit error rate. The auto-mute function can be disabled via the I²C-bus.

4. FM (AM) PATH

A high-pass filter suppresses DC offsets from the FM demodulator that may occur due to carrier Frequency offsets, and supplies the FM monitor function with DC values, e.g. for the purpose of microprocessor controlled carrier search or fine tuning functions.

An adaptive de-emphasis is available for Wegener-Panda 1 encoded satellite programs.

The de-emphasis stage offers a choice of settings for the supported TV standards.

The 2-channel decoder performs the dematrixing of $102(L + R)$, R to L and R signals of $102(L + R)$ and $102(L - R)$ to L and R signals or of channel 1 and channel 2 to L and R signals, as demanded by the different TV standards or user preferences.

Automatic FM dematrixing is also supported.

Using the high deviation mode, only channel 1 (mono) can be demodulated. The scaling is -6 dB compared to 2-channel decoding.

5. MONITOR

This function provides data words from the FM demodulator outputs and FM and NICAM signals for external use, such as carrier search or fine tuning. The peak level of these signals can also be observed. Source selection and data read out are performed via the I²C-bus.

6. DIGITAL CROSSBAR SWITCH

The input channels are derived from the FM and NICAM paths, while the output channels comprise I²S-bus and the audio DACs to the analog crossbar switch. It should be noted that there is no connection from the external analog audio inputs to the digital crossbar switch.

7. DIGITAL AUDIO OUTPUT

The digital audio output interface comprises an I²S-bus output port and a system clock output. The I²S-bus port is equipped with a level adjustment facility that can change the signal level in a ± 15 dB range in 1 dB steps. Muting is possible, too, and outputs can be disabled to improve EMC performance.

The I²S-bus output matrix provides the functions for forced mono, stereo, channel swap, channel 1 or channel 2.

Automatic selection for TV applications is possible. In this case the microcontroller program only has to provide a user controlled sound A or sound B selection.

8. STEREO CHANNEL TO THE ANALOG CROSSBAR PATH

A level adjustment function is provided with control positions of 0 dB, +3 dB, +6 dB and +9 dB in combination with the audio DACs. The Automatic Volume Level (AVL) function provides a constant output level of -20 dB (full-scale) for input levels between 0 dB (full-scale) and -26 dB (full-scale).

There are some fixed decay time constants to choose from, i.e. 2, 4 or 8 seconds.

Automatic selection for TV applications is possible. In this case the microcontroller program only has to provide a user controlled sound A or sound B selection.

9. GENERAL

The level adjustment functions can provide signal gain at multiple locations. Great care has to be taken when using gain with large input signals, e.g., due to overmodulation, in order not to exceed the maximum possible signal swing, which would cause severe signal distortion. The nominal signal level of the various signal sources to the digital crossbar switch should be 15 dB below digital full-scale (-15dB full-scale).

Description of the analog audio section

1. ANALOG CROSSBAR SWITCH AND ANALOG MATRIX

The TDA9874A has one external analog stereo input, one mono input, one 2-channel and one single-channel output port. Analog source selector switches are employed to provide the desired analog signal routing capability, which is done by the analog crossbar switch section.

The basic signal routing philosophy of the TDA9874A is that each switch handles two signal channels at the same time (e.g. left and right, language A and B) directly at the source.

Each source selector switch is followed by an analog matrix to perform further selection tasks, such as putting a signal from one input channel, say language A, to both output channels or for swapping left and right channels. The analog matrix provides the functions given in the follow table. Automatic matrixing for TV applications is also supported.

Analog matrix functions

MODE	MATRIX OUTPUT	
	L OUTPUT	R OUTPUT
1	L input	R input
2	R input	L input
3	L input	L input
4	R input	R input

The external and mono inputs accept signal levels of up to 1.4 V (RMS). By adding external series resistors to provide suitable attenuation, the external input could be used as a SCART input. Whenever the external or mono input is selected, the output of the DAC is muted to improve the crosstalk performance.

The TDA9874A comprises a 2-channel audio DAC and an additional single-channel audio DAC for feeding signals from the DSP section to the analog crossbar switch. These DACs have a resolution of 15 bits and employ four-times oversampling and noise shaping.

The output buffers provide a gain of 0 dB and offer a muting possibility. The post filter capacitors of the audio DACs are connected to the buffer outputs.

The diagram illustrates the internal architecture of the AD9058, showing the signal path from inputs to outputs. The inputs on the left are labeled: mono (AM), EXTIL, EXTIR, DACL, DACR, and DACM. These inputs feed into a 'source select' block, which contains three switch-like symbols. The outputs of the source select block feed into a 'matrix' block, which also contains three switch-like symbols. The outputs of the matrix block are connected to three output buffers, labeled OUTL, OUTR, and OUTM. The diagram shows how the source select and matrix blocks route the input signals to the appropriate output buffers.

The standby mode (see Section 7.3.3) disables most functions and reduces power dissipation of the TDA9874A. It provides no other function.

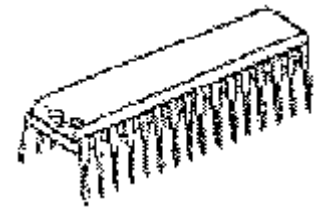
The diagram illustrates the internal architecture of the AD1885. It features three main input paths: **stereo**, **mono**, and **PS-2**. The **stereo** and **mono** inputs feed into an **ANALOG CROSSBAR SWITCH**. The **mono** input also feeds into a **DIGITAL CROSSBAR SELECT** block. The **PS-2** input feeds into a **PS-2 DECODER** block. The **ANALOG CROSSBAR SWITCH** outputs to **MATRIX** and **BUFFER** blocks, which then lead to **STEREO OUTPUT** and **MONO OUTPUT**. The **DIGITAL CROSSBAR SELECT** block outputs to a **LEVEL ADJUST** block, which then feeds into a **DACs** block. The **PS-2 DECODER** block outputs to a **MATRIX** block, which then feeds into a **LEVEL ADJUST** block, which finally outputs to the **PS-2** output.

4.NJW1136 AUDIO PROCESSOR with Subwoofer Output

■ GENERAL DESCRIPTION

THE **NJW1136** is a sound processor with subwoofer output includes all of functions processing audio signal for TV, such as tone control, balance, volume, mute, and AGC function. Also the **NJW1136** includes the LPF for subwoofer output and bass boost function. The original surround system reproduces natural surround sound and clear vocal orientation. All of internal status and variables are controlled by IIC BUS interface.

■ PACKAGE OUTLINE

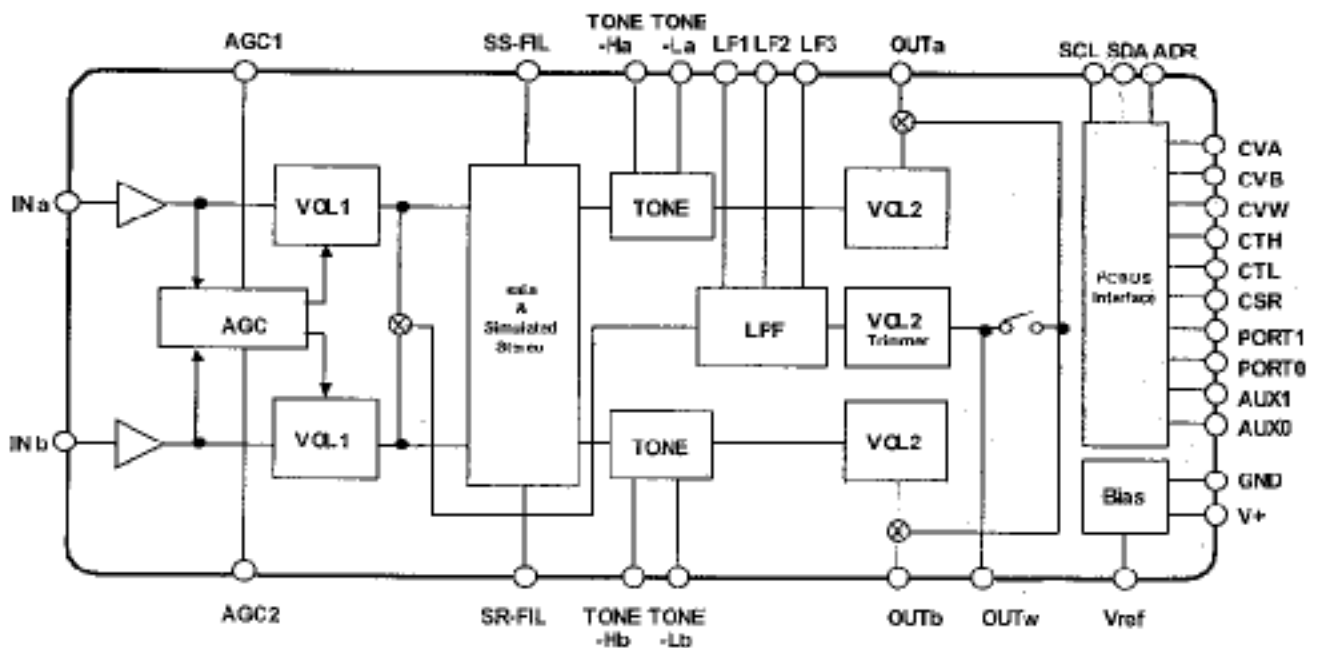


NJW1136D

■ FEATURES

- Operating Voltage: 8 to 13V
- 3ch Output(Lch, Rch, Subwoofer ch) / 2ch Output(Lch, Rch)
- LPF Filter (Adjustable cut off frequency by external parts)
- AGC Circuit (It reduces volume difference among input sources.)
Adjustable AGC boost level by external parts and AGC compression level by IIC BUS
- NJRC Original Surround System
- Simulated Stereo
- IIC BUS Interface
- Bi-CMOS Technology
- Package Outline DIP32

■ BLOCK DIAGRAM



■ PIN CONFIGURATION

No.	symbol	Function	No.	symbol	Function
1	INa	Ach input terminal	17	V+	Supply voltage terminal
2	SR-FIL	Surround filter terminal	18	Vref	Reference voltage terminal
3	SS-FIL	Simulated stereo filter terminal	19	CSR	DAC output terminal for surround control
4	TONE-Ha	Ach tone control(treble)filter terminal	20	CTL	DAC output terminal for tone control(bass)
5	TONE-La	Ach tome control(bass)filter terminal	21	CTH	DAC output terminal for tone control(treble)
6	OUTw	Subwoofer output terminal	22	CVW	Bch DAC output terminal for LPF trimmer
7	OUTa	Ach output terminal	23	CVB	Bch DAC output terminal for volume and balance
8	AGC1	AGC attack and recovery time setting terminal	24	CVA	Ach DAC output terminal for volume and balance
9	AUX0	Auxiliary 3 values voltage output terminal (0.0V,2.5V,5V)	25	AGC2	AGC boost level setting terminal
10	AUX1	Auxiliary 2 values voltage output terminal (0.0V,5V)	26	OUTb	Bch output terminal
11	PORT0	Logic input terminal	27	TONE-Lb	Bch tone control (bass) filter terminal
12	PORT1	Logic input terminal	28	TONE-Hb	Bch tone control (treble) filter terminal
13	ADR	Slave address setting terminal	29	LF3	LPF filter3 terminal
14	SDA	I ² Cdata terminal	30	LF2	LPF filter2 terminal
15	SCL	I ² C cock terminal	31	LF1	LPF filter1 terminal
16	GND	Ground terminal	32	INb	Bch input terminal

■ ABSOLUTE MAXIUM RATING (Ta=25°C)

Basic Structure

1. Internal Connections

TMPA8809 has two pieces of IC chip in one package, using Multi-Chip-Package(MCP) technology. One is a micro controller (MCU) and the other one is a signal processor (SP) for a color TV.

	Signal Name	Direction	Description
1	SCL	M to S	Internal IIC bus SCL
2	SDA	Bi-direction	Internal IIC bus SDA
3	OSD R	M to S	OSD signal connection
4	OSD G	M to S	OSD signal connection
5	OSD B	M to S	OSD signal connection
6	OSD Y/BL	M to S	OSD display control
7	OSD I, CS OUT	M to S	OSD half-tone control/Test pattern signal
8	C-Video	S to M	Composite video signal from internal video switch, for CCD
9	C-Sync	S to M	Composite sync. signal from sync. Separator, for CCD
10	HD	S to M	Horizontal timing pulse regenerated from FBP, for OSD
11	VD	S to M	Vertical timing pulse from sync. Separator, for OSD
12	CLK	M to S	8 MHz clock
13	AV _{DD}	M to S	Reference voltage for C-Video interface
14	ADC	S to M	A/D converter monitoring RF-AGC, R-Y and B-Y

There are some internal connections between these two ICs for handing below signals.
Functions of SP from MCU are controllable through the IIC bus of the internal connections.

2.Power Supply

TMPA8809 has some power supplies and GND pins. Power supplies related MCU use be applied at the first. Power supplies for H.V cc and TV D.Vcc are the second with at least 100 ms delay after MCU power ON. The other power supplies are the last, which are recommended to be supplied from a regulator a regulator circuit using FBP.

3.Crystal Resonator

TMPA8809 requires only crystal resonator, in stead that a conventional two-chip solution requires two resonators at least, one for MCU and the other one for SP. An oscillation clock with the crystal resonator of TMPA8809 is supplied for MCU operation, PIF VCO automatic alignment, alignment free AFT, of functions work properly, so that designing the oscillation frequency accurately is required. The spec of crystal is recommended to be within.

fosc:8 MHz+/-20 ppm

ftemp:8 MHz+/-40 ppm (-20°C to +65°C)

While RESET of MCU is active, the MCU function stops. Hardware and software initialization sequence including power supplies control is required, because status of any hardware after the RESET period is unknown especially horizontal oscillator which is a very basic timing generator of SP operation.

Terminal Interface

MCU Block

Pin No.	Pin Name	I/O	Function	Interface Circuit
1	P61/LED1/ ADC 8 bit (/KWU5) (AIN5) (LED1)	I/O (input) (input) (output)	Key on wake up input A/D converter analog input High current sink open drain output	
2	P60/ADC 8 bit (/KWU4) (AIN4)	I/O (input) (input)	Key on wake up input A/D converter analog input	
3	P53/ADC 8 bit/TC1/ Int2 (/KWU0) (AIN0) (TC1) (INT2)	I/O (input) (input) (input) (input)	Key on wake up input A/D converter analog input Timer/counter input External interrupt input	
4	uP DV _{SS}	Power Supply	GND	—

Pin No.	Pin Name	I/O	Function	Interface Circuit
5	/Reset	I/O	Reset signal input or watchdog timer output Address trap reset output	
6 7	XOUT XIN	Output Input	X'tal connecting pins	
8	TEST	Input	Test pin for out-going test. Be tied to low.	
9	μP DV _{DD} 5 V	Power Supply	V _{DD} Supply 5 V	
10	μP VV _{SS}	Power Supply	GND for Slicer circuit	—
54	μP MPAGND	Power Supply	GND for Oscillator circuit	—
55	μP AV _{DD} 5 V	Power Supply	V _{DD} for OSD Oscillator circuit Supply 5 V	

Pin No.	Pin Name	I/O	Function	Interface Circuit
56	P56	I/O		
57	P52/SDA (SDA)	I/O (I/O)	IIC bus serial data input/output	
58	P51/SCL (SCL)	I/O (I/O)	IIC bus serial clock input/output	

Pin No.	Pin Name	I/O	Function	Interface Circuit
59	P50/PWM 7 bit/TC2/ Int0 (/PWM8) (TC2) (/INT0)	I/O (output) (input) (input)	7-bit D/A conversion (PWM) output Timer/Counter input External interrupt input	
60	P40/PWM 14 bit (/PWM0)	I/O (output)	14/12-bit D/A conversion (PWM) output	
61	P20/Int5/ Stop (/INT5) (/STOP)	I/O (input) (input)	External interrupt input STOP mode release signal input	

Pin No.	Pin Name	I/O	Function	Interface Circuit
62	P31/Int4/ TC3 (INT4) (TC3)	I/O (input) (input)	External interrupt input Timer/Counter input	
63	P30/Int3/ RXIN (INT3) (RXIN)	I/O (input) (input)	External interrupt input Remote control signal preprocessor input	
64	P63/LED2/ (LED2)	I/O (output)	High current sink open drain output	

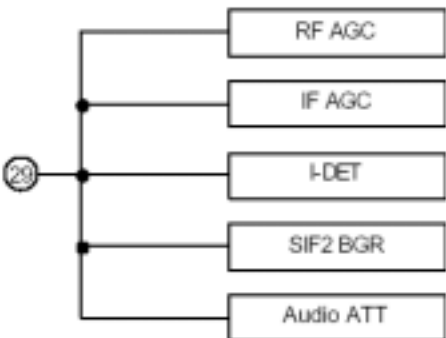
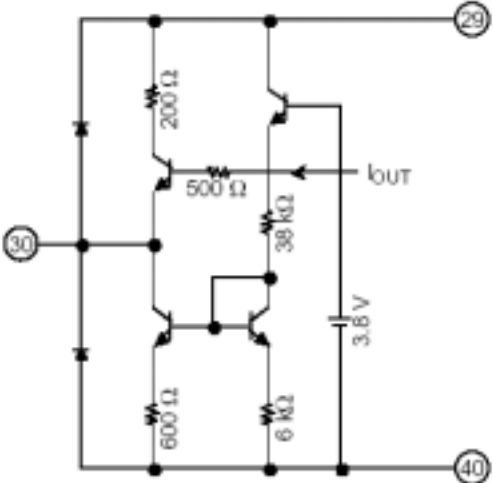
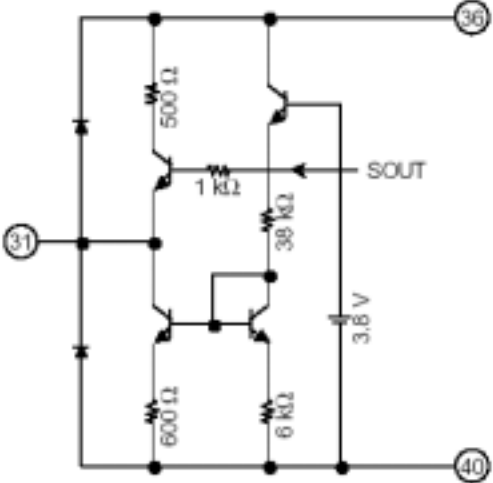
Signal Processor Block

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
11	TV AGND	GND terminal for Analog block.	—	—
12	FBP in	Input terminal for FBP.		
13	HOUT	Output terminal for Horizontal driving pulse.		
14	HAFC 1	Terminal to be connected capacitor for HAFC filter. This terminal voltage controls H VCO frequency.		

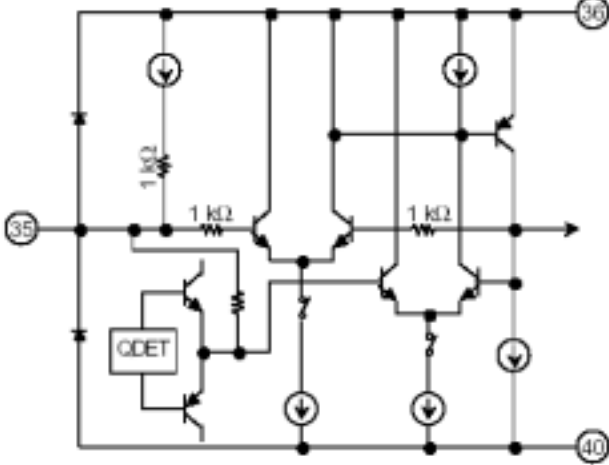
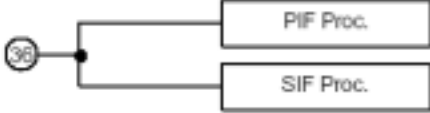
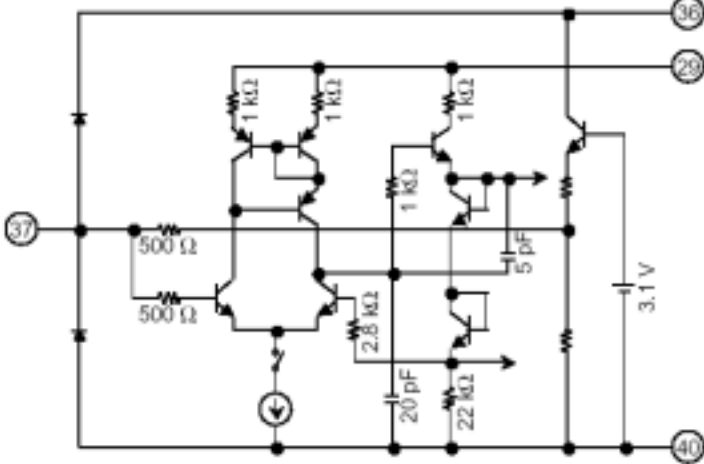
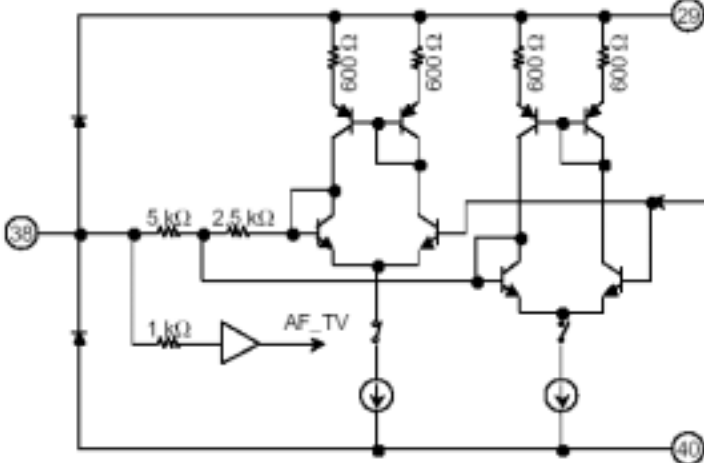
Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
15	V saw	Terminal to be connected capacitor to generate V saw signal. V saw amplitude is kept constant by V AGC function.		
16	V OUT	Output terminal for Vertical driving pulse.		
17	H.V _{CC} 9 V	V _{CC} terminal for DEF circuit. Supply 9 V.		—
18	YS in	Terminal for switching of PIP mode ----- PIP signal (YUV input) 2.5V ----- Main signal (TV / EXT / YC) ----- GND----- Ys voltage		—

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
19	Cb input	Input terminal for Cb signal.		
20	Y input	Input terminal for Y signal. (Input level = 1 Vp-p)		
21	Cr input	Input terminal for Cr signal. It is recommended that input impedance is low.		
22	TV DGND	GND terminal for Digital block.	—	—

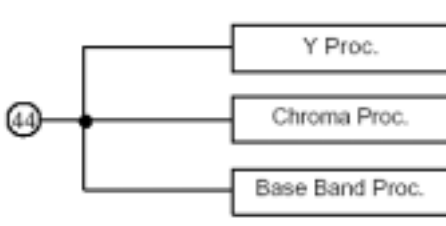
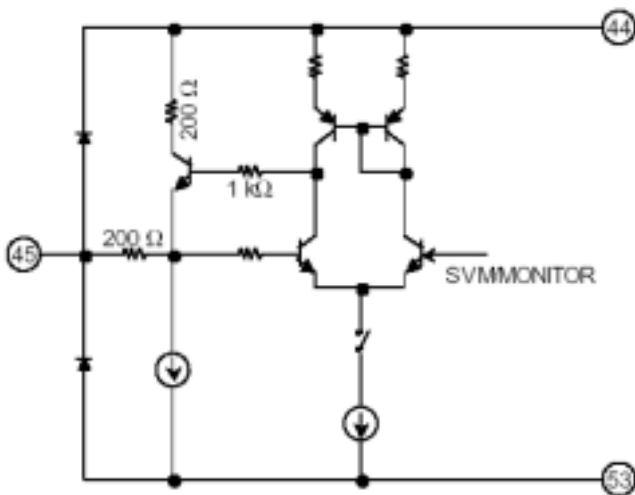
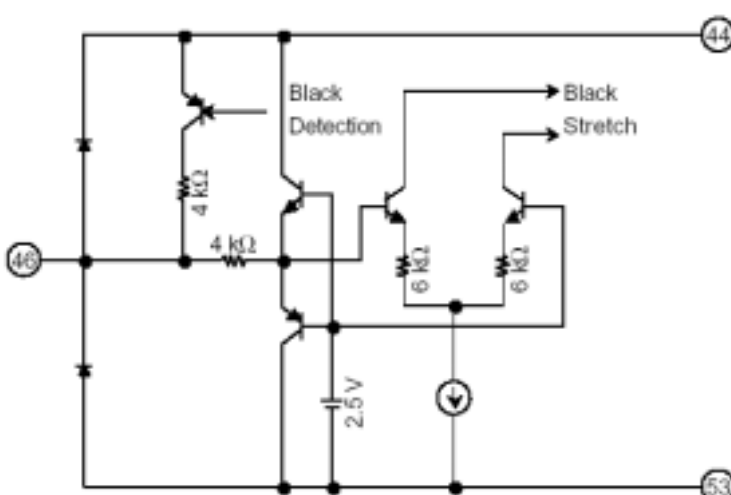
Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
23	CIN	Input terminal for Chroma signal.		
24	EXT CVBS/Y (V2 IN)	Input terminal for Video signal. (Input level = 1 Vp-p)		
25	TV DV _{CC}	V _{CC} terminal for Digital block. This terminal voltage is clipped about 3.3 V by regulator circuit. Supply TV DV _{CC} voltage from HV _{CC} (#17) voltage via 270Ω.		—

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
29	IF V _{CC} 9 V	V _{CC} terminal for IF circuit. Supply 9 V.		—
30	TVout	Output terminal for detected PIF signal. (Output level = 2.2 Vp-p)		
31	SIF out	Output terminal for detected SIF signal.		

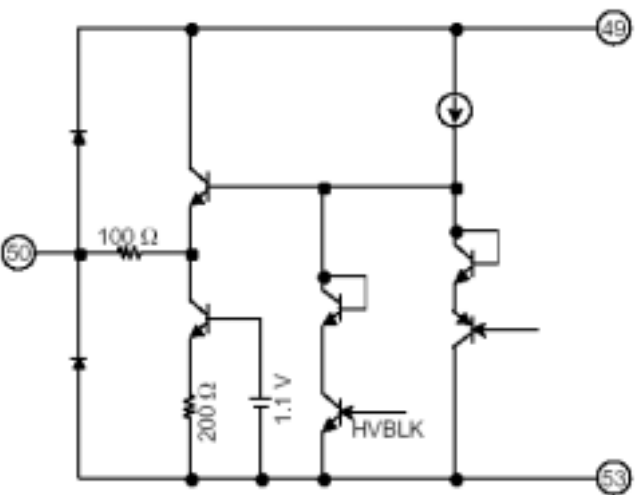
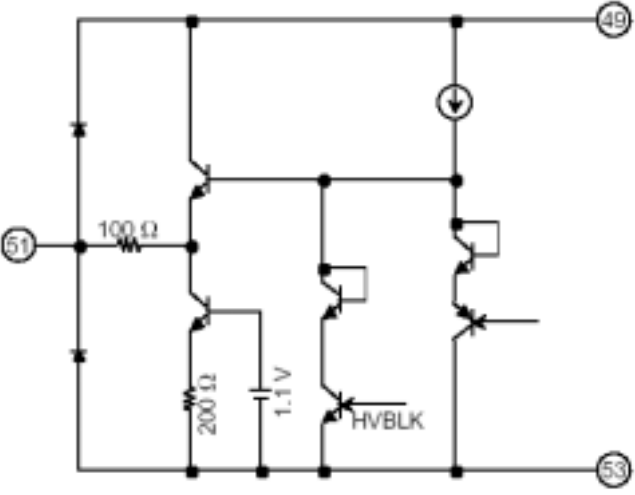
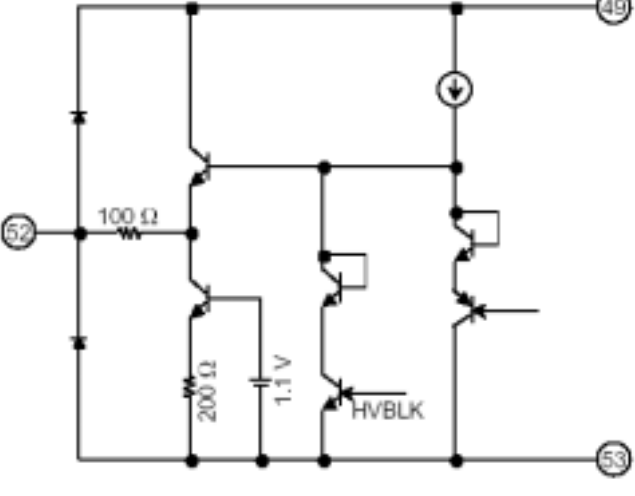
Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
32	EHT in	Input terminal for EHT feedback signal.		
33	H.correc/ SIF in	Input terminal for H correction and 2 nd SIF.		
34	DC NF	Terminal to be connected capacitor for DC Negative Feedback from SiF Det output.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
35	PIF PLL	Terminal to be connected with loop filter for PIF PLL. This terminal voltage is controlled PIF VCO frequency.		
36	IF V _{CC} 5 V	V _{CC} terminal for IF circuit. Supply 5 V.		—
37	S-Reg.F	Terminal to be connected capacitor for stabilizing internal bias.		
38	Deemph / Audi out	Terminal to be connected capacitor for SIF Det De-Emphasis.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
39	IF AGC	Terminal to be connected with IF AGC filter.		
40	IF GND	GND terminal for IF circuit.	—	—
41 42	IF IN	Input terminals for IF signals. Pin 41 and Pin 42 are both input poles of differential amplifier.		
43	RF AGC	Output terminal for RF AGC control level.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
44	YC V _{CC} 5 V	V _{CC} terminal for Y/C circuit. Supply 5 V.		—
45	Monitor out	Output terminal for CVBS or Y signal selected by BUS (video SW).		
46	Black Det	Terminal to be connected with Black Det filter for black stretch.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
47	APC Fil (Chrome PLL filter)	Terminal to be connected with APC filter for Chroma demodulation. This terminal voltage controls frequency of VCXO.		
48	IKin	Input terminal to sense AKB cathode current.		
49	RGB V _{CC} 9 V	V _{CC} terminal for RGB circuit. Supply 9 V.		-

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
50	ROUT	Output terminal for R signal.		
51	GOUT	Output terminal for G signal.		
52	BOUT	Output terminal for B signal.		
53	TV AGND	GND terminal for Analog block.	—	—

5.TMPA8809CPAN

MCU and Signal Processor for a PAL/NTSC TV

The TMPA8809CPAN is an integrated circuit for a PAL/NTSC TV. A MCU and a TV signal processor are integrated in a 64-pin shrink DIP package. The MCU contains 8-bit CPU, ROM, RAM, I/O ports, timer/counters, A/D converters, an on-screen display Controller, remote control interfaces, IIC bus interfaces and the Closed Caption decoder. The TV signal processor contains PIF, SIF, Video, multi-standard chroma, Deflection, RGB processors.

MROM: TMPA8809CPAN

OTP: TMPA8807PSAN

Features

MCU

- High speed 8-bit CPU (TLC-870/X series)
- Instruction execution time: 0.5 μ s (at 8 MHz)
- 48-Kbytes ROM, 2-Kbytes RAM
- ROM correction
- 12 I/O ports
- 14-bit PWM output 1 ch for a voltage synthesizer
- 7-bit PWM output 1 chan
- 8-bit A/D converter 3 ch for a touch-key input with key ON wake-up CIRCUIT
- Remote control signal preprocessor
- Two 16-bit in
- Two 8-bit internal timer/counter 2 ch
- Time base timer, watchdog timer
- 16 interrupt sources: external 5, internal 11
- IIC bus interface (multi-master)
- STOP and IDLE power saving modes

TV Processor IF

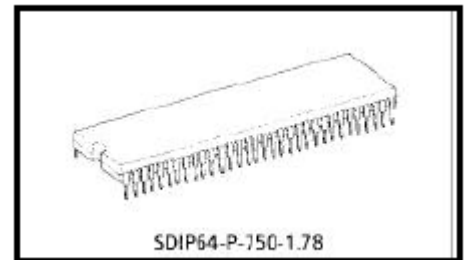
- Integrated PIF VCO aligned automatically
- Negative demodulation PIF
- Multi-frequency SIF demodulator without external Tank-coil

Video

- Integrated chroma traps
- Black stretch
- Y-gamma

Chroma

- Integrated chroma BPFs
- PAL/NTSC demodulation



Weight: 8.85 g (typ.)

CCD Decoder

- Digital data slicer for NTSC

OSD

- Clock generation for OSD display
- Front ROM characters: 384 characters
- Characters display: 32 columns $\times$ 12 lines
- Composition: 16 $\times$ 18 dots
- Size of character: 3 (line by line)
- Color of character: 8 (character by character)
- Display position: H 256/V 512 steps
- BOX function
- Fringing, smoothing, Italic, underline function
- Conform to CCD REGULATION
- Jitter elimination

RGB/Base-Band

- Integrated 1 H base-band delay line
- Base-band TINT control
- Internal OSD interface
- Half-tone and transparent for OSD
- External YcbCr interface for OSD
- RGB cut-off/drive controls by bus
- ABCL (ABL and ACL combined)

Sync.

- Integrated fH $\times$ 640 VCO
- DC coupled vert. ramp output (single)
- EW correction with EHT input

Basic Structure

2. Internal Connections

TMPA8809 has two pieces of IC chip in one package, using Multi-Chip-Package(MCP) technology. One is a micro controller (MCU) and the other one is a signal processor (SP) for a color TV. There are some internal connections between these two ICs for handing below signals.

	Signal Name	Direction	Description
1	SCL	M to S	Internal IIC bus SCL
2	SDA	Bi-direction	Internal IIC bus SDA
3	OSD R	M to S	OSD signal connection
4	OSD G	M to S	OSD signal connection
5	OSD B	M to S	OSD signal connection
6	OSD Y/BL	M to S	OSD display control
7	OSD I, CS OUT	M to S	OSD half-tone control/Test pattern signal
8	C-Video	S to M	Composite video signal from internal video switch, for CCD
9	C-Sync	S to M	Composite sync. signal from sync. Separator, for CCD
10	HD	S to M	Horizontal timing pulse regenerated from FBP, for OSD
11	VD	S to M	Vertical timing pulse from sync. Separator, for OSD
12	CLK	M to S	8 MHz clock
13	AV _{DD}	M to S	Reference voltage for C-Video interface
14	ADC	S to M	A/D converter monitoring RF-AGC, R-Y and B-Y

Functions of SP from MCU are controllable through the IIC bus of the internal connections.

2.Power Supply

TMPA8809 has some power supplies and GND pins. Power supplies related MCU use be applied at the first. Power supplies for H.V cc and TV D.Vcc are the second with at least 100 ms delay after MCU power ON. The other power supplies are the last, which are recommended to be supplied from a regulator a regulator circuit using FBP.

3.Crystal Resonator

TMPA8809 requires only crystal resonator, in stead that a conventional two-chip solution requires two resonators at least, one for MCU and the other one for SP. An oscillation clock with the crystal resonator of TMPA8809 is supplied for MCU operation, PIF VCO automatic alignment, alignment free AFT, of functions work properly, so that designing the oscillation frequency accurately is required. The spec of crystal is recommended to be within.

fosc:8 MHz+/-20 ppm

ftemp:8 MHz+/-40 ppm (-20°C to +65°C)

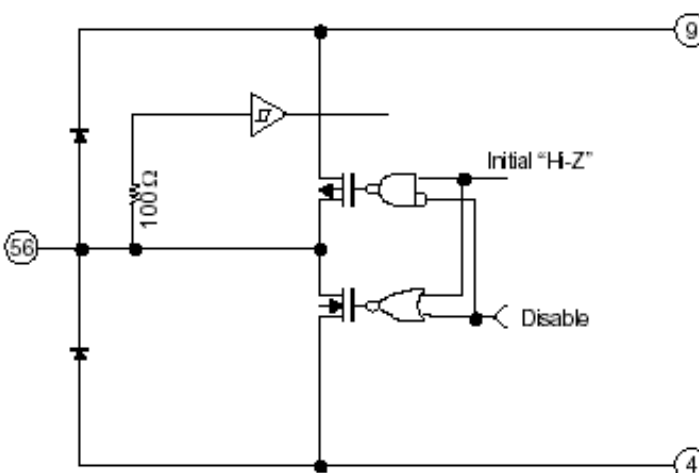
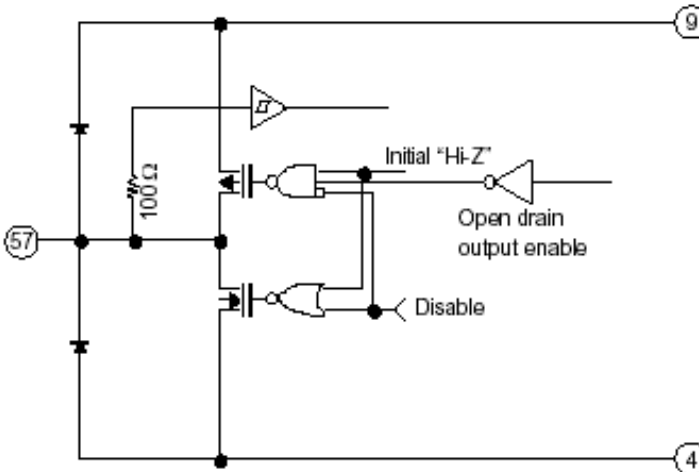
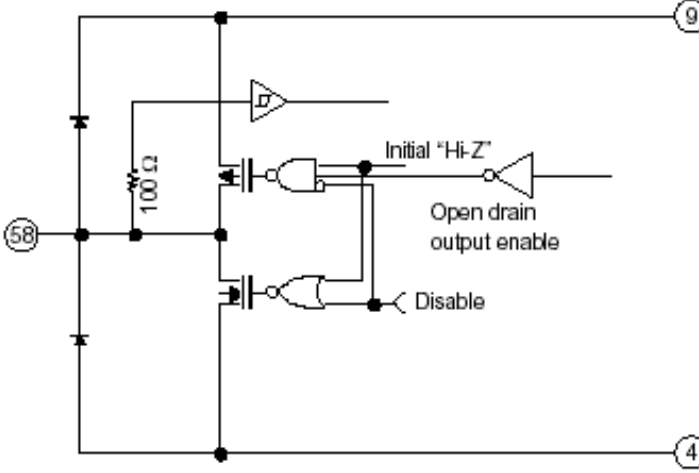
While RESET of MCU is active, the MCU function stops. Hardware and software initialization sequence including power supplies control is required, because status of any hardware after the RESET period is unknown especially horizontal oscillator which is a very basic timing generator of SP operation.

Terminal Interface

MCU Block

Pin No.	Pin Name	I/O	Function	Interface Circuit
1	P61/LED1/ ADC 8 bit (/KWU5) (AIN5) (LED1)	I/O (input) (input) (output)	Key on wake up input A/D converter analog input High current sink open drain output	
2	P60/ADC 8 bit (/KWU4) (AIN4)	I/O (input) (input)	Key on wake up input A/D converter analog input	
3	P53/ADC 8 bit/TC1/ Int2 (/KWU0) (AIN0) (TC1) (INT2)	I/O (input) (input) (input) (input)	Key on wake up input A/D converter analog input Timer/counter input External interrupt input	
4	uP DVSS	Power Supply	GND	—

Pin No.	Pin Name	I/O	Function	Interface Circuit
5	/Reset	I/O	Reset signal input or watchdog timer output Address trap reset output	
6 7	XOUT XIN	Output Input	X'tal connecting pins	
8	TEST	Input	Test pin for out-going test. Be tied to low.	
9	uP DV _{DD} 5V	Power Supply	V _{DD} Supply 5 V	
10	uP VV _{SS}	Power Supply	GND for Slicer circuit	—
54	uP MPAGND	Power Supply	GND for Oscillator circuit	—
55	uP AV _{DD} 5V	Power Supply	V _{DD} for OSD Oscillator circuit Supply 5 V	

Pin No.	Pin Name	I/O	Function	Interface Circuit
56	P56	I/O		
57	P52/SDA (SDA)	I/O (I/O)	IIC bus serial data input/output	
58	P51/SCL (SCL)	I/O (I/O)	IIC bus serial clock input/output	

Pin No.	Pin Name	I/O	Function	Interface Circuit
59	P50/PWM 7 bit/TC2/ Int0 (/PWM8) (TC2) (/INT0)	I/O (output) (input) (input)	7-bit D/A conversion (PWM) output Timer/Counter input External interrupt input	
60	P40/PWM 14 bit (/PWM0)	I/O (output)	14/12-bit D/A conversion (PWM) output	
61	P20/Int5/ Stop (/INT5) (/STOP)	I/O (input) (input)	External interrupt input STOP mode release signal input	

Pin No.	Pin Name	I/O	Function	Interface Circuit
62	P31/Int4/ TC3 (INT4) (TC3)	I/O (input) (input)	External interrupt input Timer/Counter input	
63	P30/Int3/ RXIN (INT3) (RXIN)	I/O (input) (input)	External interrupt input Remote control signal preprocessor input	
64	P63/LED2/ (LED2)	I/O (output)	High current sink open drain output	

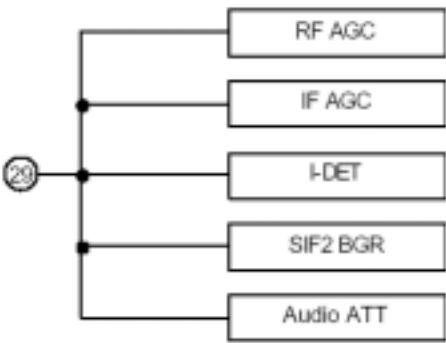
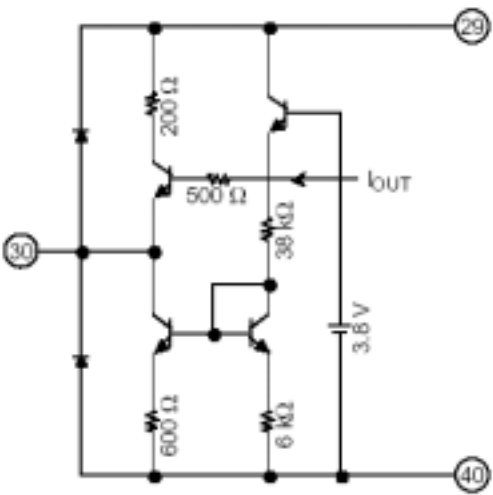
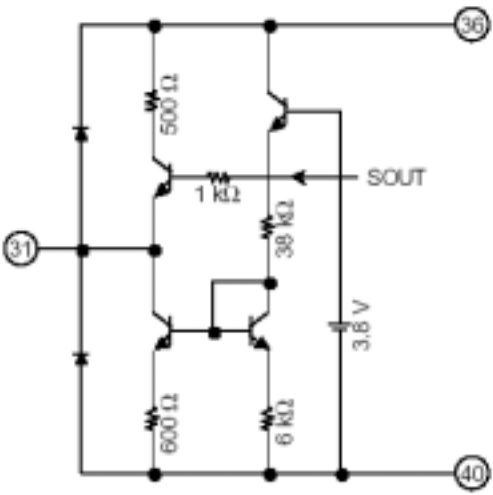
Signal Processor Block

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
11	TV AGND	GND terminal for Analog block.	—	—
12	FBP in	Input terminal for FBP.		
13	HOUT	Output terminal for Horizontal driving pulse.		
14	HAFC 1	Terminal to be connected capacitor for HAFC filter. This terminal voltage controls H VCO frequency.		

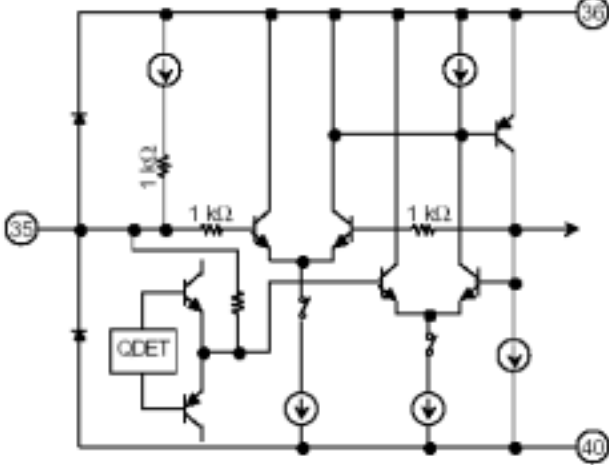
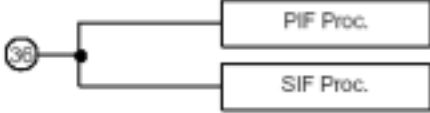
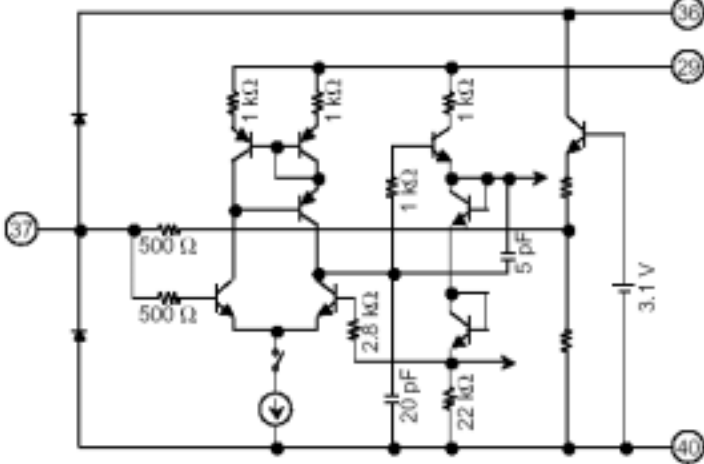
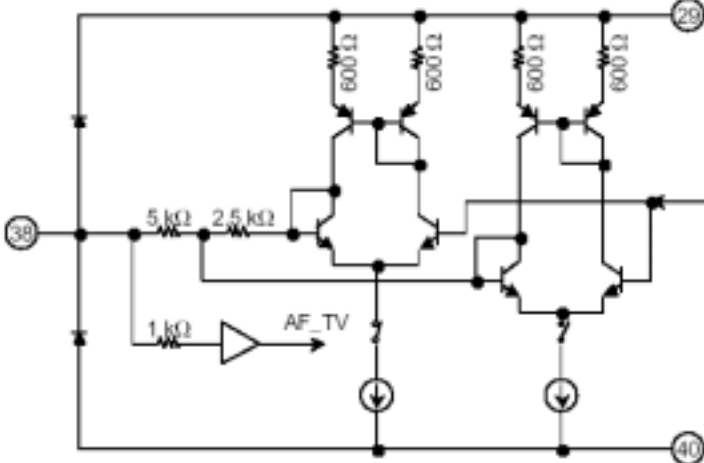
Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
15	V saw	Terminal to be connected capacitor to generate V saw signal. V saw amplitude is kept constant by V AGC function.		
16	V OUT	Output terminal for Vertical driving pulse.		
17	H.V _{CC} 9 V	V _{CC} terminal for DEF circuit. Supply 9 V.		—
18	YS in	Terminal for switching of PIP mode ----- PIP signal (YUV input) 2.5V ----- Main signal (TV / EXT / YC) ----- GND----- Ys voltage		—

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
19	Cb input	Input terminal for Cb signal.		
20	Y input	Input terminal for Y signal. (Input level = 1 Vp-p)		
21	Cr input	Input terminal for Cr signal. It is recommended that input impedance is low.		
22	TV DGND	GND terminal for Digital block.	—	—

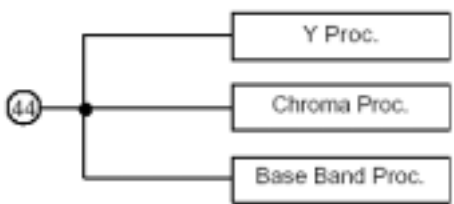
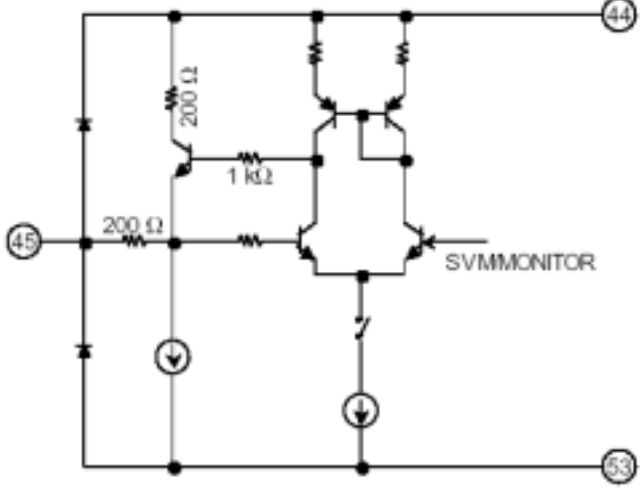
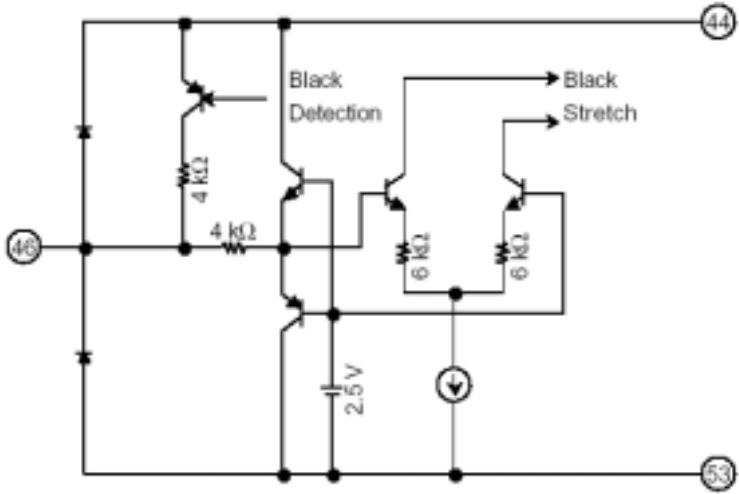
Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
23	CIN	Input terminal for Chroma signal.		
24	EXT CVBS/Y (V2 IN)	Input terminal for Video signal. (Input level = 1 Vp-p)		
25	TV DV _{CC}	V _{CC} terminal for Digital block. This terminal voltage is clipped about 3.3 V by regulator circuit. Supply TV DV _{CC} voltage from HV _{CC} (#17) voltage via 270Ω.		—

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
29	IF V _{CC} 9 V	V _{CC} terminal for IF circuit. Supply 9 V.		—
30	TVout	Output terminal for detected PIF signal. (Output level = 2.2 Vp-p)		
31	SIF out	Output terminal for detected SIF signal.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
32	EHT in	Input terminal for EHT feedback signal.		
33	H.correct/ SIF in	Input terminal for H correction and 2 nd SIF.		
34	DC NF	Terminal to be connected capacitor for DC Negative Feedback from SIF Det output.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
35	PIF PLL	Terminal to be connected with loop filter for PIF PLL. This terminal voltage is controlled PIF VCO frequency.		
36	IF V _{CC} 5 V	V _{CC} terminal for IF circuit. Supply 5 V.		—
37	S-Reg.F	Terminal to be connected capacitor for stabilizing internal bias.		
38	Deemph / Audi out	Terminal to be connected capacitor for SIF Det De-Emphasis.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
39	IF AGC	Terminal to be connected with IF AGC filter.		
40	IF GND	GND terminal for IF circuit.	—	—
41 42	IF IN	Input terminals for IF signals. Pin 41 and Pin 42 are both input poles of differential amplifier.		
43	RF AGC	Output terminal for RF AGC control level.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
44	YC V _{CC} 5 V	V _{CC} terminal for Y/C circuit. Supply 5 V.		—
45	Monitor out	Output terminal for CVBS or Y signal selected by BUS (video SW).		
46	Black Det	Terminal to be connected with Black Det filter for black stretch.		

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
47	APC Fil (Chrome PLL filter)	Terminal to be connected with APC filter for Chroma demodulation. This terminal voltage controls frequency of VCXO.		
48	IKin	Input terminal to sense AKB cathode current.		
49	RGB V _{CC} 9 V	V _{CC} terminal for RGB circuit. Supply 9 V.		-

Pin No.	Pin Name	Function	Interface Circuit	I/O Signal
50	ROUT	Output terminal for R signal.		
51	GOUT	Output terminal for G signal.		
52	BOUT	Output terminal for B signal.		
53	TV AGND	GND terminal for Analog block.	—	—

PART V. Adjusting Description

1.TMPA8827+M113 software adjustment specification(SAMPLING MODEL:2918AE)

M113 chassis is used as I²C bus control chassis for 25 inch or bigger TV in our company ,use TOSHIBA TMPA8809 TWO IN ONE TV process chip, and software M113 designed by R&D dept. ourselves. First production runing please use flash edition TOSHIBA TMPA8827 to be programmable, then, change to OTP edition TOSHIBA TMPA8827, take care of this point in your production. Secondly, adopting I²C bus structure, we can use remote handset to accomplish adjustment, also automatic adjustment instruments can be used in some adjustment items, including pincushion distortion 、 H-width、 IF VCO can be adjustment by remote handset or automatic adjustment instruments. Of course both the voltage of screen and focus will be adjusted by hands. (remark : if it has NICAM function , the adjustment way of this module with TDA9874APS can be your reference.)

To solve the problems happened in production process, workers who join to assemble this model should master its specifications. This model TV set have two modes: customer mode, factory mode. The former is setted for customer ; the latter is for production in factory and repairing in aftersales. Customer mode can be operated with remote handset or keys in front panel, but factory mode only be operated with remote handset.

The method to enter factory mode is as below: after power on, press“ volume▼ ”on front panel, waiting for the scale to “ 0 ”, then press key “ DISPLAY ” on the remote handset, there will be “ D ” displayed on the screen shows it enters in factory mode. Now the shortcut keys on the factory adjustment remote handset can be (remark : customer mode remote handset also can be used, it can be used in repairing adjustment, press key“ STANDBY ” to exit factory mode, but system setting data “ 6 ” should be kept the state before storing, do not to change it.) changed factory data and restored in memory. For factory production, you can set “ BIT-0 ” to “ 1 ” in menu “ OPT ”, so everytime you press the remote handset can enter “ D-MODE ” directly after open the sets in AC or DC power sets on, also it can be activated by key “ D-MODE ” on the remote handset, but do not forget to set“ BIT-0 ”to“ 0 ”in menu“ OPT ”and power them on in AC or DC mode one time in the end of production, so the sets will disable factory mode.

The key “ I²C ” on the remote handset can interrupt the communication CPU with I²C bus, usually this interrupt state can be used in automatic whitebalance adjustment and auto geometry distortion adjustment.

No .	Adjustment items	Adjusted part	recommend position	Input signal /mothed condition	Setting method
1	Screen voltaage	VR Screen on FBT		All pattern "IRGB cutoff"should be set to 80 , there maybe different IRGB-cutoff settings with different CRTs)	Pre key "mute",make vertical to stop scan,then there will be a center horizontal line,adjust this VR to make the horizontal line can just been seen(minimum visible intensity)
2	Fucus voltage	VR Focus on FBT		Signal:cross hatch, input port:AV&TV	Adjust VR focus,observe the center and four sides of the picture until the horizontal and vertical lines become clear

3	Key 1 (PAL)	HIT (V-SIZE) VP50 (V-POS) VLIN (V-LINE) VSC (VS-CORRECT) VBLK (V-UP-DOWN MASK) VCEN (V-CENTER) OSDH (OSD POS)	11 02 00 01 00 11 1F	SIGNAL:cross hatch INPUT PORT:unlimited PAL SYSTEM	Check items in sequence with remote handset, observe the picture until the cross hatch in the center and all sides/corners are in accord V-POS:adjust vertical-center in the middle center of the picture OSDH:adjust OSD(50HZ) position suitable, commonly according to commend value. VCEN, VBLK:adjust according to commend value
4	Key 1 (NTSC)	HIT S (V-SIZE) VP60 (V-POS) VLIS (V-LINE) VSS (VS-CORRET) VBLK (V-UP-DOWN BLANK) VCEN (V-CENTER) OSDHS (OSD POS)	14 00 02 02 00 11 1D	SIGNAL:CR- OSS HATCH INPUT PORT:UNLIMITED NTSC SYSTEM	Check items in sequence with remote handset, observe the picture until the cross hatch in the center and all sides/corners are in accord, v-pos adjustment:make vertical-center in the center of the picture OSDHS adjustment:adjust OSD(60HZ) position is suitable, commonly according to comment value. VCEN, VBLK according to commend value.
5	Key 2 (PAL)	HPOS (H-CENTER) DPC (H-PINCUSHION CORRCTION) KEY (TRAPZOID) WID (H-SIZE) ECCT (TOP CONNER CORRCTION) ECCB (BOTTOM CONNER CORRCTION) VEHT (V-CURRENT BEAM) HEHT (H-CURRENT BEAM)	11 26 15 2A 0D 0C 04 07	SIGNAL:RECTAGAL WHIT/BLACK BACKGROUND INPUT PORT:UNLIMITED SYSTEM : PAL	Check items in sequence with remote handset, observe the picture until the cross hatch in the center and all sides/corners are in accord HPOS data:horizontal center in the center of the picture WID data:H-size suitable. VEHTS and HEHTS is for the picture size stability when changing the brightness of the screen.Receive pattern of cross hatch with black background and then change to white background, then compare the vertical and horizontal size between black and white background.adjust VEHT and HEHT until you get the minimum difference of screen size.After you adjust VEHT and HEHT, you must re-adjust vertical and horizontal size.

6	KEY 2 (NTSC)	HPS (H-CENTER) DPCS (PINCUSHION COR) KEYS (TRAPZOID COR) WIDS (H-SIZE) ECCTS (TOP CORNNER CORRECT) ECCBS (BOTTOM CORNNER CORRECT) VEHTS (V-CURRENT BEAM) HEHTS (V-CURRENT BEAM)	15 21 10 29 12 11 04 07	SIGNAL:CROSS HATCH WHITE/BLACKBACKG ROUND INPUT PORT:UNLIMITED SYSTEM:NTSC	Check items in sequence with remote handset,observe the picture until the vertical lines in center/corner picture are straight HPS: adjust horizontal center in the middle center WIDS:adjust H-width properly. Adjust VEHT and HEHT using same method of PAL system.Also need readjustment of vertical and horizontal Size.
7	DIGITAL KEY "3" (STATIC ADJUST)	CNTX CNTN BRTX BRTN COLX COLN TNTX TNTN	5A 05 20 1D 3F 00 42 28	SIGNALGREY SCALE/ HALF-COLOR BAR INPUT PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the recommend values
8	DIGITAL KEY "4" (STATIC ADJUST)	BRTC COLC COLS COLP SCOL SCNT CNTC TNTC	37 57 47 F0 04 0F 40 48	SIGNAL:GREY SCALE/ HALF-COLOR BAR INPUT PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the commend values
9	DIGITAL KEY "5" (HIGH DEFINITI ON ADJUST)	ST3 SV3 ST4 SV4 SVD ASSH SHPX SHPN	20 20 20 20 15 04 3F 1A	SIGNAL:MULTIBURST PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the commend values
10	DIGITAL KEY 7 (STATIC ADJUST)	RFAGC SBY SRY BRTS TXCS RGCN SECD MUTT STAT	26 0A 08 0D 1F 00 08 70 30	INPUT PORT:TV SIGNAL:LADDER AND HALF COLOR BAR LEVEL : 60dB	Adjust TXCS to 1F Adjust RGCN to 00 RFAGC:adjust noise point in the picture just been seen minimum intensity SBY and SRY:in the models have SECAM function,adjust colorbar to best tinge BRTS :adjust the sceond grey bar just to been seen(minimum visible intensity) in "standard" displayed in the picture

11	DIGITAL KEY "8" (SOUND STATE ADJUST , SOUND CURVE ADJUST)	V01 VOLUME WHICH SET ON 1 V25 VOLUME WHICH SET ON 1 V50 VOLUME WHICH SET ON 1 V100 VOLUME WHICH SET ON 100 BASC BASX TREC WOFC AVC	3A B0 DC FF 40 72 40 39 0E	SOUND SIGNAL:1KHz AV INPUT	1)adjust the volume scale to"100",using voltage level meter and oscillator,watching speaker level,input 5rms signal。 Adjust "v100" according to the requirement of power and distortion,record the level value Vm (dB) 2)adjust volume scale to"1",adjust " V01 ", make speaker sound level less 67dB than Vm。 3)adjust volume scale to"25", adjust " V25 ", make speaker sound level less 24dB than Vm。 4)adjust volume scale to" 50", adjust " V50 ", make speaker sound level less 10dB than Vm。 remark : always adjust sound curve to commend value , using remote handset adjust any other items in sequence until to commend value.
12	DIGITAL KEY"8" (SOUND EFFECTIVE STATE ADJUST B.E/WOOFE R CURVE ADJUST	NEWS SPACES NEWT SPACET WOFF B01 B25 B50	14 5A 14 5A 00 4F 68 7F	Enrich sound signal AV input	Adjust these items to the commend values
13	CALENDAR KEY (OTHER ADJUSTMENT)	SVM SVM1 SVM2 SVM3 PYNX PYNN PYXS PYNS	03 02 03 00 28 15 22 04	SIGNAL:UNLIMITED INPUT PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the commend values
14	NOTE KEY (OTHER ADJUSTMENT)	CLTO CLTM CLVO CLVD ABL DCBS DEF	4B 4C 4D 48 27 33 01	SIGNAL:UNLIMITED INPUT PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the commend values

15	GAME KEY (OTHER ADJUSTMENT)	OSD1 OSDF1 OSD2 OSDF2 HAFC NOIS UCOM	30 66 48 75 09 01 00	SIGNAL:UNLIMITED INPUT PORT:UNLIMITED SYSTEM:UNLIMITED	Adjust these items to the commend values
16	KEY 0 (COLOR TEMPERATURE)	R B (RED CUTOFF) G B (GREEN CUTOFF) B B (BLUE CUTOFF) G D (GREEN GAIN) B D (BLUE GAIN)	80 80 80 40 40	SIGNAL:UNLIMITED SIGNAL: LADDER SYSTEM:PAL	

In the white balance adjustment, options R-cutoff,B-cutoff,G-cutoff used to adjust low brightness whitebalance,and options G-gain,B-gain used to adjust high brightness white balance.

Adjustment without instructions :

- ①first adjust the data of G-cutoff to "50"
- ②adjust BLUE data
- ③press V-scan to stop vertical scan.
- ④observe the color of the center horizontal line.
- ⑤back to the normal V-scan
- ⑥repeat step②、③、④、⑤until the color of center horizontal line change to amethyst.
- ⑦adjust G-cutoff data
- ⑧repeat steps③、④、⑤、⑦until the center horizontal line change to white.
- ⑨observe ladder signal,adjust G-gain and B-gain let the most bright bar change to purity white.

remark : 1. press key "channel" will let R/G/B cutoff and R/G/B gain appear in circle.

2. the adjusted items by automatic instruments as sample to what of instruments system.

2.M113 series parameter setting

OPT

BIT	Store in EEPROM menu 6
0	D-MODE : 0 : disable, 1 : enable.
1	0 : No SYNC signal in TV, mute disable 1 : No SYNC signal in TV, mute able.
2	0 : Picture mute when change channel 1: picture mute disable when change channel.
3	M system 25KHz AUDIO GAIN : 0 : 927mV, 1:500mV
4	NO SYNC signal: 0:NO AFT 1:AFT
5	0: AV change , mute disable ; 1: AV change , mute .
6	KOREA (PAL50Hz picture mute): 0:NO 1: YES
7	PIN64 STANDBY: 0: HIGH LEVEL STANDBY ; 1: LOW LEVEL STANDBY

FLG0

BIT	Store in EEPROM menu 6	TMPA8827
0	PIF OVER MOD. : 0 : nomal, 1:enable	
1	N BUZZ CANCEL: 0: disable 1 : enable	
2	Outscreen eliminate brightpoint: 0: disable 1 : enable	
3	NC	
4	When hotel mode abled 00 : enter memory status when power on	
5	1x:enter TV status when power on.01: enter AV status when power on.	
6	Hotel mode : 0: disable 1 : enable	
7	Change channel VCO adjust: 0:enable 1 : disable	

FLG1

BIT	Store in menu 6	
0	0: Auto search (NO.7 KEY) disable ; 1 : Auto search (NO.7 KEY) enable	
1	SECAM: 0: disable 1 : able	
2	LOGO: 0: disable 1 : able	
3	TINT phase polarity 0: positive 1: negative	
4,5,6	IF setting ; 001:45.75MHz, 011:38.8MHz, 100:38MHz	
7	APC 0:APC=000 1:3.58 NTSC and 4.43 ,APC=011 other systems: 000	TMP8827 24H D5—D7

STBY

BIT	Store in EEPROM menu 6
0	When outscreen eliminate brightpoint, standby level starting time data byte× 4 0 μ m.
1	
2	
3	
4	00 : Enter standby status after power on, 01 : electrify and power on .
5	10 or 11:After power on enter in the status of the time before power off
6	Auto sleep function 0:disable 1:able
7,	NC
HD_DELAY	Start after standby,H-off time(outscreen eliminate brightpoint) data× 30 μ m

MODE 0

BIT	MODE 0.Store in EPPROM menu 6
0	NICAM 0 : able 1 : disable
1	English 0 : able 1 : disable
2	Russian : 0 : disable 1 : disable
3	Vietnamese : 0 : able 1 : disable
4	Picture MUTE type 0:Y mute 1:RGB mute
5	Default system after auto search 100: BG 101:I 110:DK 111:M 0XX: the
6	same as customer setup (X shows 1 or 0)
7	

MODE 1

BIT	MODE 1 .Store in EEPROM menu 6
0	BG 0 : OFF, 1 : ON
1	I 0 : OFF, 1 : ON
2	DK 0 : OFF, 1 : ON
3	M 0 : OFF, 1 : ON
4	VIDEO2 0 : OFF, 1 : ON
5	VIDEO3 0 : OFF, 1 : ON
6	YUV 0 : OFF, 1 : ON
7	NC

MODE2

BIT	OPT : Operation system. Store in EEPROM menu 6	
0	NC	
1	Scart mode 0 : disable 1 : able	
2	0 : B.E, 1 : SuperWoofer	
3	NC	
4	NC	
5	NC	
6	NC	
7	NC	

DEF: Store in EEPROM menu NOTE

DEF	Vertical AGC reference 0 : depend on YC power source 1:depend on inside power source	TMPA8827 1DH D4
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CLVD、CLTO、CLTM、CLVO

BIT	Store in EEPROM menu NOTE	Corresponding
0	Y delay 000 : -40ns 001 :0ns 010:40ns	15H From D0 to D2
1	011: 80ns 011: 120ns 101: 160ns	
2	110: 200ns 111: 240ns	
3	NTSC matrix 00: NTSC 93 ⁰ (Japan) 01:NTSC108 ⁰ (USA)	03H From D6 to D7
4	10: 110 ⁰ DVD 11:90 ⁰ DVD	
5	C GAMMA 0: off 1: on	02H D7
6	PAL/NTSC Eliminate color identify sensitive	17H D4

	0:1.2/1.5mVp-p 1:6.6/6.4mVp-p	
7	Eliminate color switch 0:normal 1: disable	17H D3

CLVD:YUV mode CLTO:audio system is not M under TV mode CLTM:audio system is M under TV mode

CLVO:Mode under VIDEL mode

ABL

BIT	ABL setting mode ,store in EEPROM menu NOTE	TMPA8827
0	ABL gain 00 : -0.17V 01:-0.35V 10:-0.50V 11 ; -0.65V	16H D4,D5
1		
2	ABL starting controlled point 00 : 0V 01:-0.15V 10:-0.28V 11 ; -0.38V	16H D6,D7
3		
4	YPL 0: brightness signal peak value limited 1 : brightness peak value unlimited.	00H D7
5	OSD ABL 0: able 1 : disable	16H D3
6	NC	
7	NC	

DCBS

BIT	VIDEO DATA SETUP store in EEPROM menu NOTE	TMPA8827
0	Black level extend 00:NC 01 : 25IRE 10 : 33IRE 11 : 43IRE	16H D3,D4
1		
2	Y GAMMA 00:OFF 01:90IRE 10: 82IRE 11:75IRE	15H D5,D6
3		
4	OSD level 00:96IRE 01:60IRE 10: 70IRE 11:80IRE	16H D0,D1
5		
6	NC	
7	Disappear switch 0 : inside 1:outside	0CH D7

HAFC:

BIT	Store in EEPROM menu GAME				TMPA8827
0	TV HAFC: NOIS.2=1 ,TV HAFC reference VIEDO HAFC				
1	C:NOIS.2=0, NOIS DET>NOIS the values of the 0,1bit HAFC=01 NOIS DET≤NOIS the values of the 0,1bit HAFC=00				
2, 3	VIDEO HAFC		BLACKING PERIOD	PICTURE PERIOD	1CH D4,D5
		00	3	1	
		01	4/3	1/3	
		10	2	1	
		11	OFF	OFF	

NOIS :

BIT	GAME
0	NOIS reference data setup
1	
2	HAFC value : 0 : compare value 1 : fixed value

UCOM : MCU data, Can't adjust to the bit value " 00 " . (Store in EEPROM menu GAME)

AVC:

BIT	Store in EEPROM menu 8	Corresponding to NJW1136L
0	AVC level: 00 : 100mVrms 01 : 100mVrms 10 : 100mVrms 11 : 100mVrms	05H D1,D2
1		
2	AVC: 0 : disable 1 : able	05H D0

PART VI. Troubleshooting

Flow chart

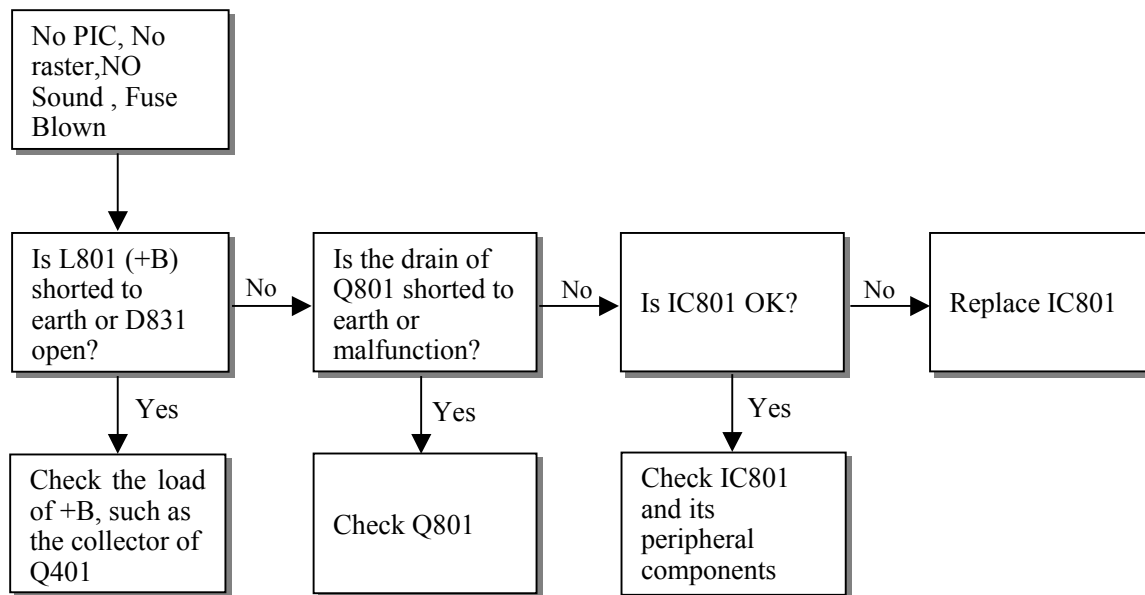


Fig.1 No picture, no raster, no sound, Fuse Blown

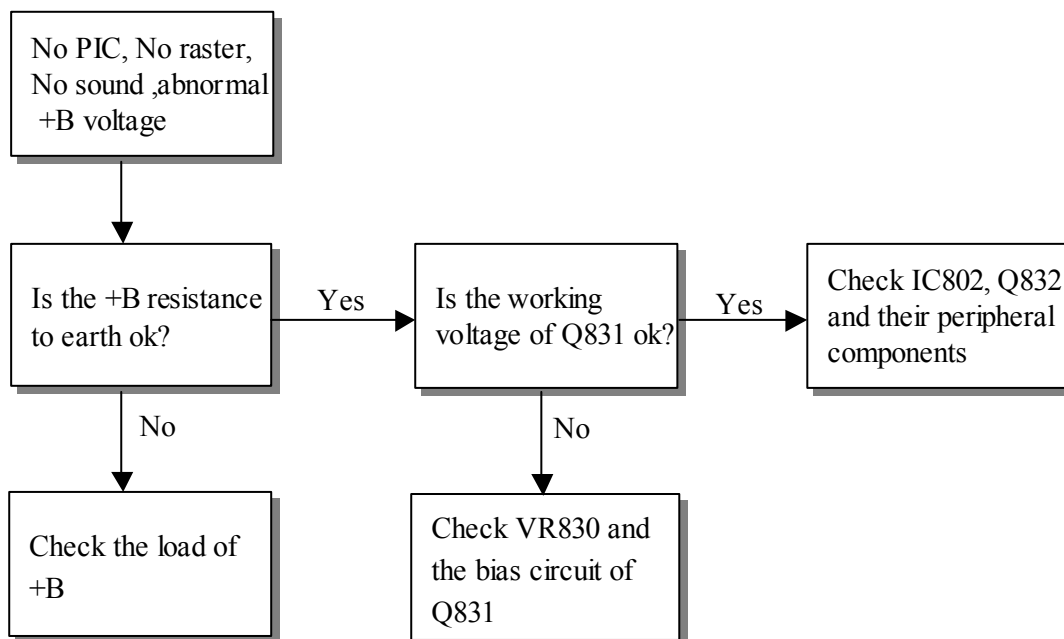


Fig.2 No picture, no raster, no sound, abnormal +B voltage

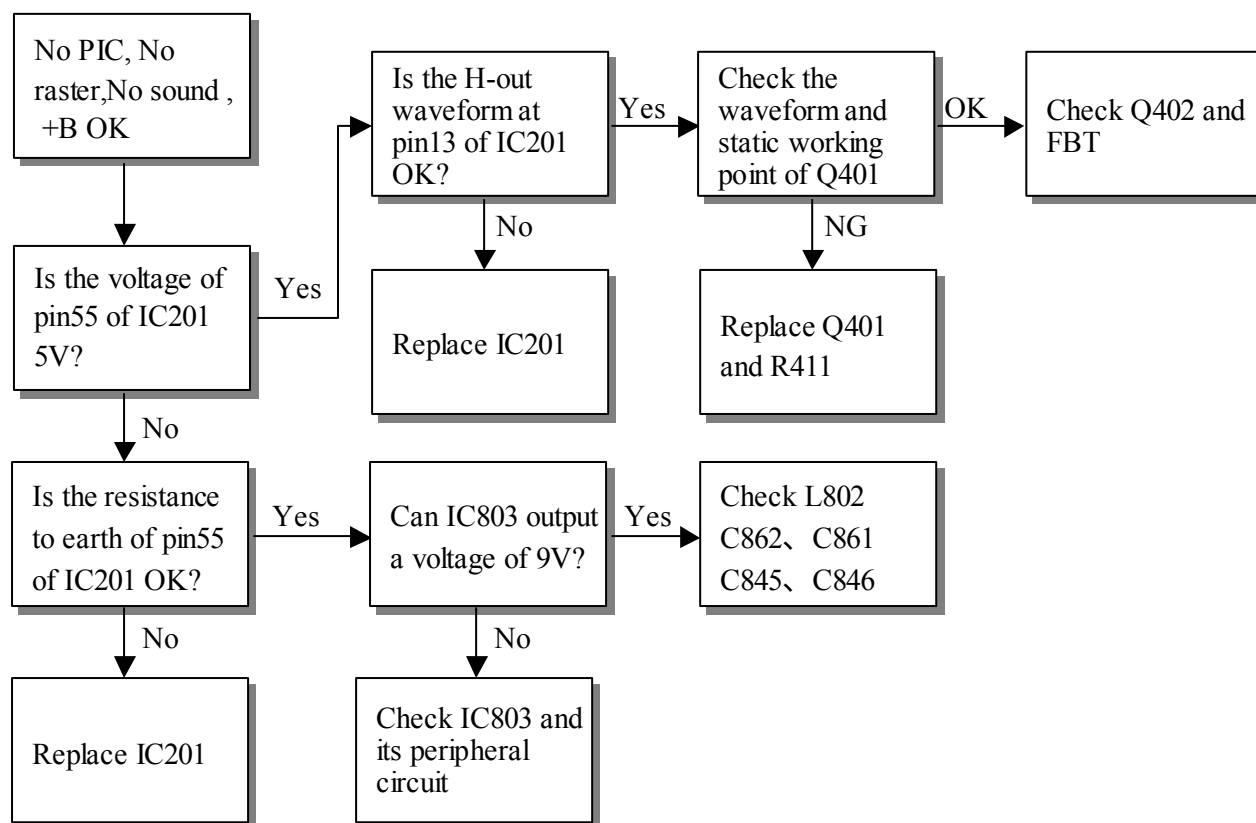


Fig.3 No picture, no raster, no sound, +B OK

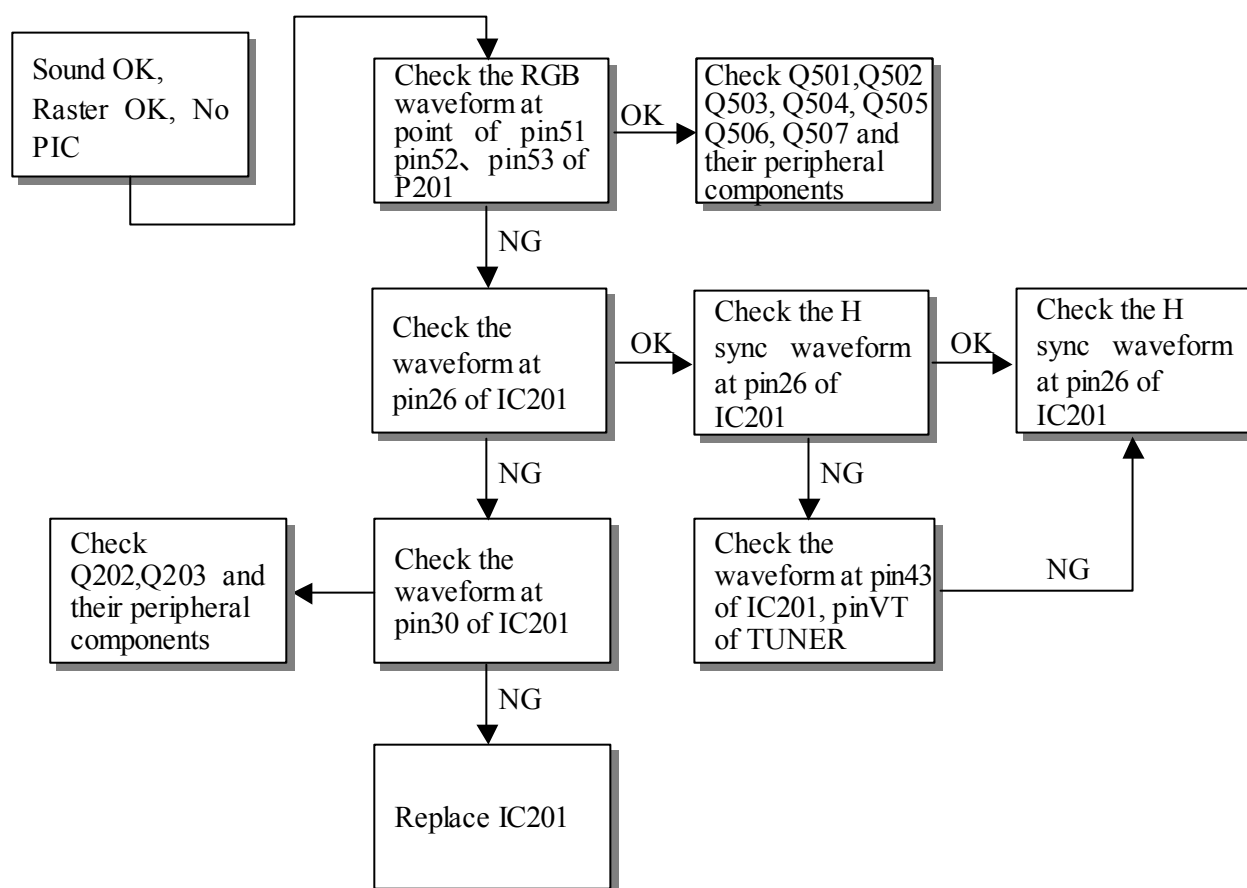


Fig.4 No picture, raster OK, sound OK

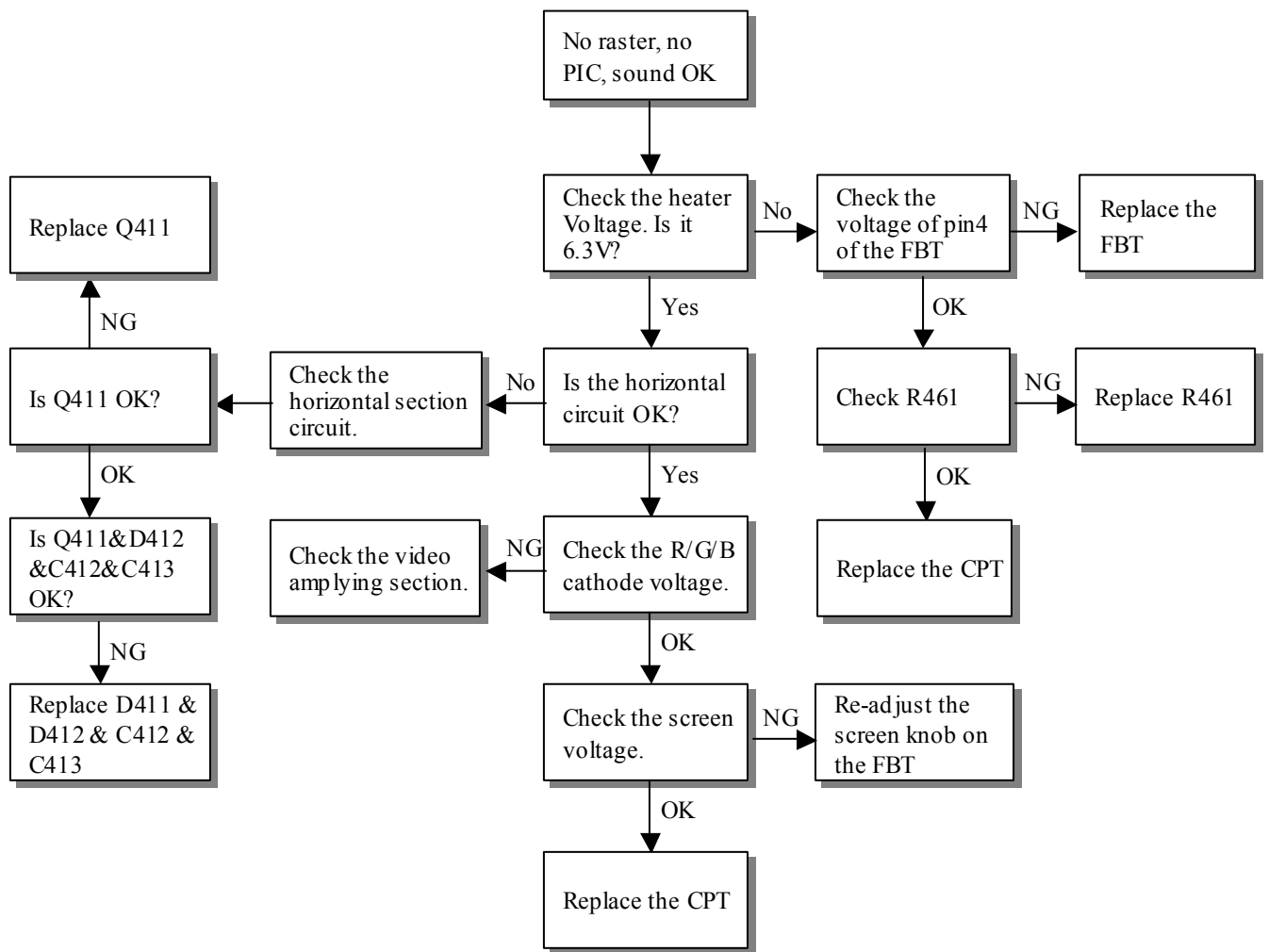
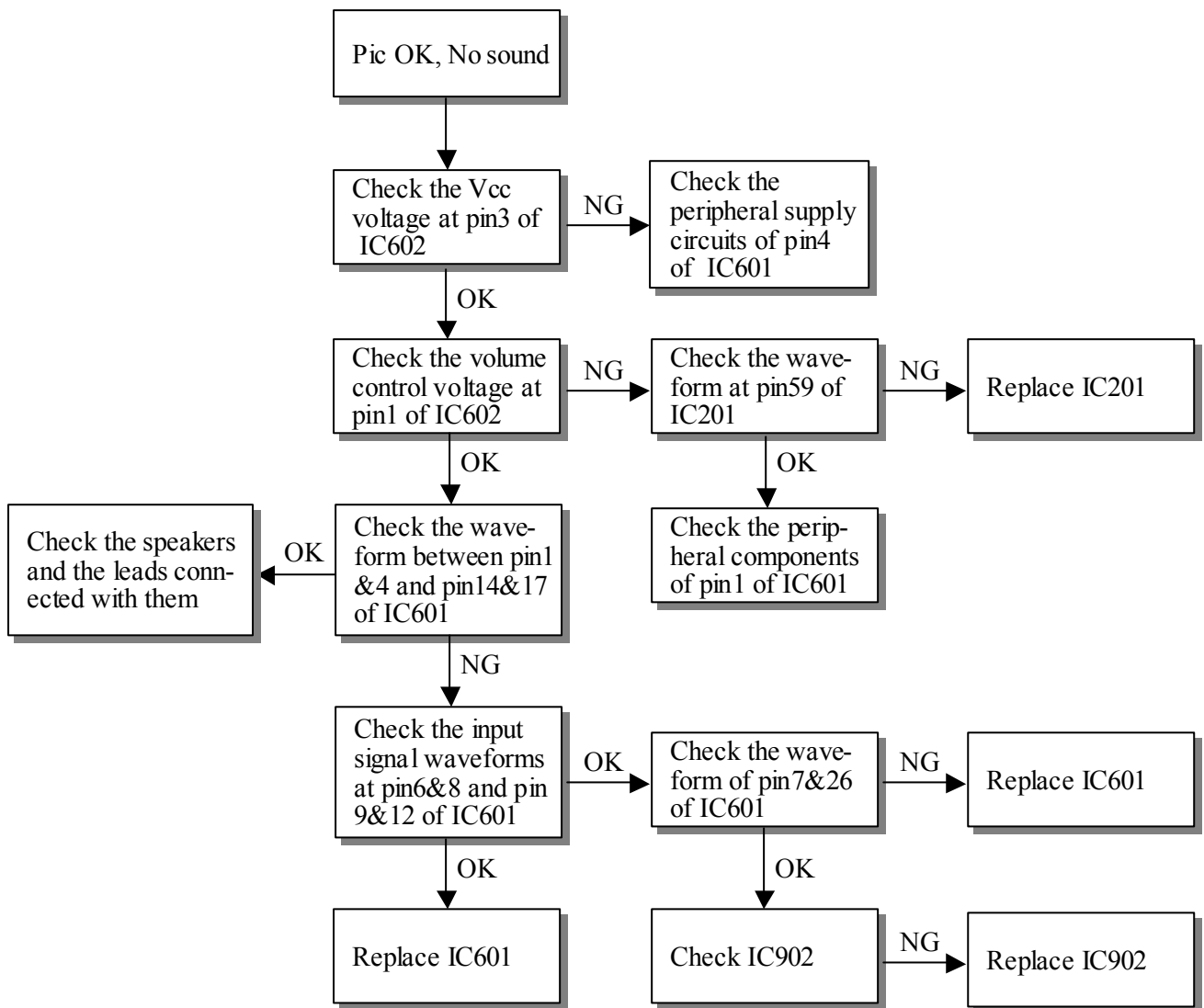
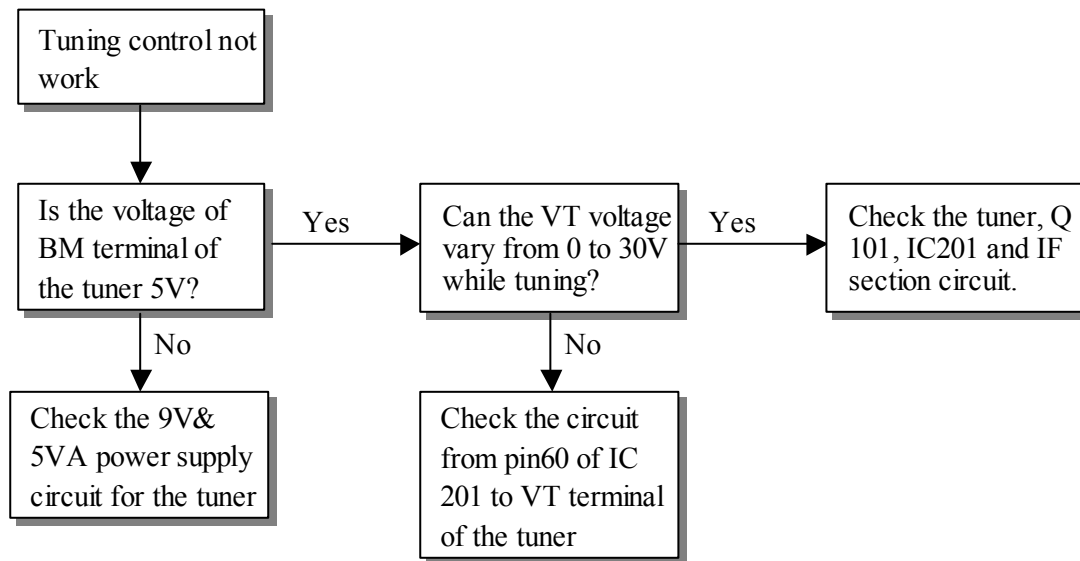


Fig.5 No raster, no picture, sound OK

**Fig.6 Picture OK, no sound**

**Fig.7 Tuning control not work**

PART VII. BOM (PREPARE BOM, ONLY FOR YOUR REFERENCE)

CMPITM	ITMDSC	QTY	REMARK
08-2918AE-CRY	ASS'Y - CRT BD	1	
10-1N4148-ABX	DIODE 1N4148 (SWITCHING)	2	D501 D502
11-A562TM-0BX	TRANSISTOR 2SA562TM-0	1	Q507
11-SC1815-YBX	TRANSISTOR 2SC1815Y	3	Q502 Q504 Q506
11-SC4544-0AX	TRANSISTOR 2SC4544	3	Q501 Q503 Q505
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	1	R515
18-CB0272-JNX	RES. C.F. 2.7K OHM 1/6W +/-5%	1	R514
18-CB0561-JNX	RES. C.F. 560 OHM 1/6W +/-5%	3	R501 R505 R510
18-CB0681-JNX	RES. C.F. 680 OHM 1/6W +/-5%	8	R502 R506 R509 R522 R503 R508 R513 R511
18-FE0272-JNX	RES. M.O. 2.7K OHM 1/2W +/-5%	3	R519 R520 R521
18-FG0183-JHX	RES. M.O. 18K OHM 2W +/-5%	3	R516 R517 R518
25-BLA100-M1X	CAP. ELEC 10 UF 250V +/-20%	1	C504
26-AMK102-KRX	CAP. CER 1000 pF 2KV +/-10% R	1	C505
26-EBP103-ZFX	CAP. CER 0.01UF 50V +80/-20% F	2	C508 C509
26-EBP391-JCX	CAP. CER. 390PF 50V +/-5% CH	2	C501 C502
26-EBP471-JCX	CAP. CER 470PF 50V +/-5% CH	1	C503
34-R220K2-1BX	COIL CHOKE 22 UH +/-10%	1	L503
35-237250-00X	FERR. BEAD HF70	2	FOR C508 (L505 & L506)
40-02501C-CRG	P.C.B. CRT BD	1	
41-WJ0050-B00	WIRE BARE JUMPER 5MM	1	L504
41-WJ0060-B00	WIRE BARE JUMPER 6mm	3	R504 R507 R512
41-WJ0075-B00	WIRE BARE JUMPER 7.5MM	1	J503
46-10967W-01X	PIN BASE *1 TJC1-1A	1	P501 FOR CRT GROUNDING HOUS.
46-30615H-04X	HS 4P24 460 F/W TJC3-4Y/SCN-4	1	P502 FOR M.BD P421
46-37030H-05X	HS 5P 2468#24 450 TJC3-5Y/SCN	1	P503 FOR M.BD P201
10-1N4148-ABX	DIODE 1N4148 (SWITCHING)	1	L501("1" TO P501"1")
41-WJ0050-B00	WIRE BARE JUMPER 5MM	1	L502
25-BCB471-M1X	CAP. ELEC 470 UF 16V +/-20%	1	C506
47-CRT004-XX0	CRT SOCKET GZS10-2-108	1	S501
08-2918AE-FBY	ASS'Y - FRONT CTL BD	1	
18-CB0103-JNX	RES. C.F. 10K OHM 1/6W +/-5%	2	R087 R081
18-CB0152-JNX	RES. C.F. 1.5K OHM 1/6W +/-5%	1	R082
18-CB0182-JNX	RES. C.F. 1.8K OHM 1/6W +/-5%	1	R083
18-CB0272-JNX	RES. C.F. 2.7K OHM 1/6W +/-5%	1	R084
18-CB0432-JNX	RES. C.F. 4.3K OHM 1/6W +/-5%	1	R085
18-CB0622-JNX	RES. C.F. 6.2K OHM 1/6W +/-5%	1	R086
46-28559W-02X	PIN BASE *2 TJC1-2A	1	P801 FOR POWER CORD
46-28559W-02X	PIN BASE *2 TJC1-2A	1	P802
46-32269H-02X	HS 2P22 200 TJC1-2Y/TJC1-2Y	1	FOR F.CTL BD(P802) & M.BD(S801)
48-256460-0C0	SWITCH POWER TV8(PCB MOUNTING)	1	SW801
48-TAC002-XX0	TACT SWITCH	7	S001 S002 S003 S004 S005 S006 S007
62-310430-0HA	LED HOLDER (COMMON)	1	
63-W30100-AB4	S/T SCREW W 3 X 10 AB	3	MTG F.CTL BD & M.BD BKT
11-IRR001-1X0	IR RECEIVER MODULE HS0038A2	1	IR001
14-LED05R-XX1	LED RED FB205	1	D080
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	1	R090
18-CB0103-JNX	RES. C.F. 10K OHM 1/6W +/-5%	1	R091
18-CB0470-JNX	RES. C.F. 47 OHM 1/6W +/-5%	1	R089
41-WJ0050-B00	WIRE BARE JUMPER 5MM	3	J081 J082 J083
11-SA1015-YBX	TRANSISTOR 2SA1015Y	1	Q081
40-2918MT-FBB	P.C.B. FRONT CTL BD	1	
46-27777H-05X	NOTICE:THIS P/N HAD BEEN CANCELED	1	P005 FOR 46-13541W-057
63-W30100-AB4	S/T SCREW W 3 X 10 AB	1	MTG F.CTL BD & M.BD BKT
51-DC0243-0CH01	POWER CORD VDE PLUG W/HOUSING	1	
25-BCB470-M1X	CAP. ELEC 47 UF 16V +/-20%	1	C080

11-C124ES-0BX	TRANSISTOR PDTC124ES (NPN)	1	Q082
08-2918AE-FCN	ASS'Y - FRONT CABINET	1	
02-GND029-XX0	ASS'Y-CRT GND WIRE & HOUS.(29)	1	
36-294960-017	DEGAUSSING COIL XC-74A-3100-TCL 29"	1	
42-81608F-XX0	SPEAKER (80MMX160MM) 8 OHM 8W	2	SP601 SP602
44-290FLW-TS4A	CRT A68AGA20X99	1	CRT01
46-26514H-04X	HS 4P A/B 500/13 RBGW TJC1-4Y	1	P411H
46-27580H-02X	HS 2P22 600/7 TJC3-2Y	2	P602H P601H
54-113970-0U0	PVC TUBE #5 L=ROLL	0.32	FOR SPK CONNECTORS
54-205140-000	SPACER CRT MOUNTING T=2MM	8	MTG CRT & F.CAB
54-310140-000	SPONGE 50 X 20MM (BKK)	2	FOR TEWTER HOLES
55-2918FC-0CNAB	FRONT CABINET	1	
55-2918FP-0HA9A	FRONT PANEL	1	
56-2918FB-0HAAB	PUSH BUTTON	1	
56-2918LE-0HCAA	LENS	1	
56-2918PK-0HAAB	POWER KNOB	1	
58-346720-1UIAA	INLAY SIDE AV	1	
59-130460-00X	RUBBER PAD (25mmX7mm)	2	STICK ON F.CAB(FOOTING)
59-312160-000	SPONGE CUSHION	8	FOR SPK & F.CAB
62-10654X-00F	UNI - TIE (2.5mmX95mm)	13	
62-216340-0UA	HOLDER POWER CORD	1	
62-262660-0HA	POWER SW. ADAPTER	1	
63-H60300-AB4	S/T SCREW H 6 X 30 AB	4	MTG CRT
63-P30060-OB3	S/T SCREW P 3 X 6 OB	1	MTG LENS & FRONT PANEL
63-W30100-AB4	S/T SCREW W 3 X 10 AB	2	MTG PUSH BUTTON & F.CAB
63-W30100-AB4	S/T SCREW W 3 X 10 AB	2	MTG SIDE AV BD & F.CAB
63-W30140-HS4	S/T SCREW W 3 X 14 HS	4	MTG SPK & F.CAB
65-A75200-20E	WASHER 7.5mmX20mmX2mm MTG CRT	4	MTG CRT
67-216790-0E0	SPRING CRT 6MMX50MMX0.6MM	2	
67-250450-0E0	SPRING KNOB	1	
67-2960LG-2A0AA	LOGO	1	
54-114000-00X	FELT TAPE (150mmX19mmX0.3mm)	3	
54-314740-0X0	CRT FIBRE SHEET (22mmX22mmX0.8mm)	4	MTG CRT TO F.CAB
54-314740-0X0	CRT FIBRE SHEET (22mmX22mmX0.8mm)	4	
08-2918AE-MAN	ASS'Y - MAIN BD	1	
07-380VI5-NX1	TUNER TELE4-801A (IEC)	1	TU101
10-0BY228-F0X	DIODE BY228	1	D411
10-0FR104-FBX	DIODE FR104	4	D412 D421 D441 D431
10-1N4001-EBX	DIODE 1N4001 (RECTIFIER)	1	D301
10-1N4148-ABX	DIODE 1N4148 (SWITCHING)	8	D201 D203 D601 D602 D603 D202 D204 D205
10-79C5V1-DBX	DIODE ZENER 5V1 1/2W 5%	1	D001
10-79C8V2-DBX	DIODE ZENER 8V2 1/2W 5%	1	D303
11-2N3904-0A1	TRANSISTOR 2N3904	1	Q004
11-C124ES-0BX	TRANSISTOR PDTC124ES (NPN)	2	Q211
11-C124ES-0BX	TRANSISTOR PDTC124ES (NPN)	1	Q602
11-SA1015-YBX	TRANSISTOR 2SA1015Y	8	Q008 Q215 Q412 Q413 Q601 Q912 Q915 Q916
11-SC1815-YBX	TRANSISTOR 2SC1815Y	7	Q003 Q007 Q201 Q202 Q203 Q204 Q205
11-SC1815-YBX	TRANSISTOR 2SC1815Y	7	Q212 Q451 Q910 Q911 Q913 Q914 Q909
11-SC2482-0BX	TRANSISTOR 2SC2482	1	Q401
11-SC3779-DBX	TRANSISTOR 2SC3779D (RF AMPL)	1	Q101
11-SD2539-0AX	TRANSISTOR 2SD2539 (HORIZ O/P)	1	Q411
11-SK2541-0BX	TRANSISTOR 2SK2541 (N-CHANNEL)	1	Q210
13-000040-52P	IC 4052	2	IC901 IC902
13-0TDA81-72S	IC TDA8172	1	IC301
13-00M24C-08P	IC EEPROM 8K M24C08	1	IC001
13-NJW113-6LP	IC NJW1136L	1	IC601
13-TDA894-4JS	IC TDA8944J	1	IC602
18-CB0100-JNX	RES. C.F. 10 OHM 1/6W +/-5%	1	R204
18-CB0101-JNX	RES. C.F. 100 OHM 1/6W +/-5%	6	R206 R207 R213 R218 R930 R203

18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	8	R003 R004 R023 R031 R032 R106 R453 R904
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	8	R906 R910 R912 R923 R924 R925 R926 R931
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	5	R944 R950 R950A R951 R951A
18-CB0562-JNX	RES. C.F. 5.6K OHM 1/6W +/-5%	2	R001 R002
18-CB0103-JNX	RES. C.F. 10K OHM 1/6W +/-5%	8	R006 R025 R026 R033 R035 R043 R227 R246
18-CB0103-JNX	RES. C.F. 10K OHM 1/6W +/-5%	7	R314 R618 R607 R605 R054 R224 R305
18-CB0104-JNX	RES. C.F. 100K OHM 1/6W +/-5%	7	R034 R210 R619 R943 R235 R236 R907
18-CB0121-JNX	RES. C.F. 120 OHM 1/6W +/-5%	1	R217
18-CB0123-JNX	RES. C.F. 12K OHM 1/6W +/-5%	2	R937 R413
18-CB0151-JNX	RES. C.F. 150 OHM 1/6W +/-5%	3	R102 R108 R228
18-CB0153-JNX	RES. C.F. 15K OHM 1/6W +/-5%	3	R933 R009 R021
18-CB0183-JNX	RES. C.F. 18K OHM 1/6W +/-5%	1	R432
18-CB0202-JNX	RES. C.F. 2K OHM 1/6W +/-5%	1	R412
18-CB0221-JNX	RES. C.F. 220 OHM 1/6W +/-5%	4	R401 R928 R929 R250
18-CB0222-JNX	RES. C.F. 2.2K OHM 1/6W +/-5%	8	R615 R903 R909 R945 R946 R946A R423 R417
18-CB0223-JNX	RES. C.F. 22K OHM 1/6W +/-5%	8	R005 R022 R208A R221 R223 R616 R905 R911
18-CB0223-JNX	RES. C.F. 22K OHM 1/6W +/-5%	7	R913 R932 R938 R939 R942 R940 R941
18-CB0224-JNX	RES. C.F. 220K OHM 1/6W +/-5%	1	R240
18-CB0271-JNX	RES. C.F. 270 OHM 1/6W +/-5%	4	R243 R244 R245 R215
18-CB0303-JNX	RES. C.F. 30K OHM 1/6W +/-5%	1	R241
18-CB0331-JNX	RES. C.F. 330 OHM 1/6W +/-5%	4	R201 R214 R608 R609
18-CB0332-JNX	RES. C.F. 3.3k OHM 1/6W +/-5%	3	R212 R216 R253
18-CB0333-JNX	RES. C.F. 33K OHM 1/6W +/-5%	4	R020 R306 R617 R225
18-CB0362-JNX	RES. C.F. 3.6K OHM 1/6W +/-5%	1	R302
18-CB0392-JNX	RES. C.F. 3.9K OHM 1/6W +/-5%	4	R007 R008 R203A R303
18-CB0394-JNX	RES. C.F. 390K OHM 1/6W +/-5%	1	R935
18-CB0470-JNX	RES. C.F. 47 OHM 1/6W +/-5%	2	R024 R927
18-CB0471-JNX	RES. C.F. 470 OHM 1/6W +/-5%	6	R036 R042 R044 R103 R208 R211
18-CB0472-JNX	RES. C.F. 4.7K OHM 1/6W +/-5%	4	R010 R315 R420 R416
18-CB0473-JNX	RES. C.F. 47K OHM 1/6W +/-5%	4	R101 R226 R610 R219
18-CB0513-JNX	RES. C.F. 51K OHM 1/6W +/-5%	1	R220
18-CB0560-JNX	RES. C.F. 56 OHM 1/6W +/-5%	1	R107
18-CB0681-JNX	RES. C.F. 680 OHM 1/6W +/-5%	1	R934
18-CB0683-JNX	RES. C.F. 68K OHM 1/6W +/-5%	2	R936 R418
18-CB0750-JNX	RES. C.F. 75 OHM 1/6W +/-5%	1	R922
18-CB0820-JNX	RES. C.F. 82 OHM 1/6W +/-5%	3	R901 R902 R908
18-CB0821-JNX	RES. C.F. 820 OHM 1/6W +/-5%	1	R307
18-CB0822-JNX	RES. C.F. 8.2K OHM 1/6W +/-5%	2	R202 R256
18-CD0479-JNX	RES. C.F. 4.7 OHM 1/4W +/-5%	1	R030
18-CD0829-JNX	RES. C.F. 8.2 OHM 1/4W +/-5%	2	R620 R613
18-CE0109-JNX	RES. C.F. 1 OHM 1/2W +/-5%	1	R304
18-CE0271-JNX	RES. C.F. 270 OHM 1/2W +/-5%	1	R209
18-EG0109-JHX	RES. FUS. 1 OHM 2W +/-5%	4	R301 R422 R431 R461
18-FE0121-JNX	RES. M.O. 120 OHM 1/2W +/-5%	1	R310
18-FE0159-JNX	RES. M.O. 1.5 OHM 1/2W +/-5%	1	R309
18-FE0472-JNX	RES. M.O. 4.7K OHM 1/2W +/-5%	1	R316
18-FF0221-JGX	RES. M.O.F 220 OHM 1W +/-5%	1	R311
18-FF0242-JGX	RES. M.O. 2.4K OHM 1W +/-5%	1	R402
18-FF0681-JGX	RES. M.O. 680 OHM 1W +/-5%	1	R419
18-FG0103-JHX	RES. M.O. 10K OHM 2W +/-5%	1	R421
18-FG0228-JHX	RES. M.O. 0.22 OHM 2W +/-5%	1	R614
18-FG0399-JHX	RES. M.O. 3.9 OHM 2W +/-5%	1	R415
18-GM0302-KTX	RES. CEMENT 3K OHM 7W +/-10%	1	R404
25-264610-M1X		1	C415
25-BBB331-M1X	CAP. ELEC 330 UF 10V +/-20%	1	C630
25-BJB479-M1X	CAP. ELEC 4.7 UF 160V +/-20%	1	C401
25-BJG101-M1X	CAP. ELEC 100 UF 160V +/-20%	1	C422
25-BKG100-M1X	CAP. ELEC 10 UF 200V +/-10%	1	C421
25-BLF100-M11	CAP. ELEC 10 UF 250V +/-20%	1	C442

26-AIC332-KBX	CAP. CER 3300 PF 500V +/-10% B	1	C402
26-AIC391-KBX	CAP. CER 390 PF 500V +/-10% B	1	C403
26-AIE221-KB1	CAP. CER 220PF 500V +/-10% B	2	C441 C431
26-EBP100-JCX	CAP. CER 10PF 50V +/-5% CH	1	C303
26-EBP101-JCX	CAP. CER 100PF 50V +/-5% CH	1	C240
26-EBP102-KBX	CAP. CER 1000 PF 50V +/-10% B	5	C226 C109 C610 C651 C003
26-EBP103-ZFX	CAP. CER 0.01UF 50V +80/-20% F	7	C007 C022 C029 C234 C238 C244 C250
26-EBP103-ZFX	CAP. CER 0.01UF 50V +80/-20% F	7	C921 C931A C922 C923 C034 C106 C107
26-EBP103-ZFX	CAP. CER 0.01UF 50V +80/-20% F	6	C108 C214 C220 C225 C418 C913
26-EBP104-ZFX	CAP. CER 0.1UF 50V +80%/-20%	5	C004 C230 C232 C236 C622
26-EBP104-ZFX	CAP. CER 0.1UF 50V +80%/-20%	4	C611 C613 C624 C643
26-EBP152-KBX	CAP. CER 1500pF 50V +/-10% B	1	C002
26-EBP181-JCX	CAP. CER 180PF 50V +/-5% CH	1	C919
26-EBP220-JCX	CAP. CER 22PF 50V +/-5% CH	1	C252
26-EBP221-JCX	CAP. CER 220PF 50V +/-5% CH	4	C006 C009 C027 C030
26-EBP222-KBX	CAP. CER 2200PF 50V +/-10% B	3	C217 C231 C306
26-EBP270-JCX	CAP. CER 27PF 50V +/-5% CH	1	C008
26-EBP330-JCX	CAP. CER 33PF 50V +/-5% CH	1	C916
26-EBP331-JCX	CAP. CER 330PF 50V +/-5% CH	3	C020 C042 C043
26-EBP390-JCX	CAP. CER 39PF 50V +/-5% CH	2	C031 C032
26-EBP471-JCX	CAP. CER 470PF 50V +/-5% CH	1	C239
27-ALQ272-J0X	CAP. M.PP 0.0027UF 1.6KV +/-5%	1	C411
27-ALQ722-J0X	CAP. M.PP 7200 pF 1.6KV +/-5%	1	C414
27-ALQ762-J0X	CAP. M.PP 7600 PF 1.6KV +/-5%	1	C412
27-MBC103-J0X	CAP. M.P.E 0.01UF 63V +/-5%	2	C626 C628
27-MBC104-J0X	CAP. M.P.E 0.1 UF 63V +/-5%	6	C625 C633 C102 C251 C243 C242
27-MBC104-J0X	CAP. M.P.E 0.1 UF 63V +/-5%	6	C241 C602 C603 C605 C652 C253
27-MBC222-J0X	CAP. M.P.E 0.0022UF 63V +/-5%	3	C653 C604 C654
27-MBC333-J0X	CAP. M.P.E 0.033UF 63V +/-5%	1	C656
27-MBC334-J0X	CAP. M.P.E 0.33UF 63V +/-5%	1	C608
27-MBC393-J0X	CAP. M.P.E 0.039UF 63V +/-5%	1	C309
27-MBC473-J0X	CAP. M.P.E 0.047 UF 63V +/-5%	3	C629 C627 C655
27-MCB272-J0X	CAP. M.P.E 2700 PF 100V +/-5%	1	C311
27-PBC224-J0X	CAP. P.E 0.22UF 63V +/-5%	2	C005 C305
27-PBC822-J0X	CAP. P.E 0.0082UF 63V +/-5%	1	C247
27-RHQ223-J0X	CAP. PP 0.022 UF 400V +/-5%	1	C413
27-RHQ563-J0X	CAP. PP 0.056 UF 400V +/-5%	1	C451
34-A109K0-1IX	COIL CHOKE 1 UH +/-10%	1	L102
34-R100J2-0EX	COIL PL - 10 UH +/-5%	2	L002 L208
34-R220J2-0EX	COIL PL - 22 UH +/-5%	3	L201 L202 L206
34-R270J2-0EX	COIL PL - 27 UH +/-5%	1	L207
34-R470J2-0EX	COIL PL - 47 UH +/-5%	1	L101
34-R829J2-0EX	COIL PL - 8.2 UH +/-5%	1	L203
36-LIN150-XX1	COIL LINEARITY 15 UH	1	L414
37-BSC290-1650X	FLYBACK BSC29-0165	1	T461
41-BF0015-3BB	WIRE UL 1007 #24 15MM	1	L411
41-WJ0065-B00	WIRE BARE JUMPER 6.5MM	1	D209
41-WJ0050-B00	WIRE BARE JUMPER 5MM	2	C215 J610
41-WJ0055-B00	WIRE BARE JUMPER 5.5 MM	1	J915
41-WJ0060-B00	WIRE BARE JUMPER 6mm	8	J004 J005 J218 J229 J301 J302 J306 J307
41-WJ0060-B00	WIRE BARE JUMPER 6mm	5	J614 J628 J909 J910 J911
41-WJ0060-B00	WIRE BARE JUMPER 6mm	4	J923 JP901 JP902 J419
41-WJ0065-B00	WIRE BARE JUMPER 6.5MM	6	J008 J023 J201 J202 J217 J602
41-WJ0065-B00	WIRE BARE JUMPER 6.5MM	6	J606 J629 J924 J925 J305 JP904
41-WJ0070-B00	WIRE BARE JUMPER 7MM	3	J919 J922 J926
41-WJ0075-B00	WIRE BARE JUMPER 7.5MM	6	J208 J210 J221 J222 J308 J413
41-WJ0075-B00	WIRE BARE JUMPER 7.5MM	6	J605 J613 J622 J623 J624 J917
41-WJ0080-B00	WIRE BARE JUMPER 8 MM	3	J006 J603 J604
41-WJ0085-B00	WIRE BARE JUMPER 8.5MM	7	J007 J025 J223 J608 J627 J902 J907

41-WJ0090-B00	WIRE BARE JUMPER 9MM	1	J024
41-WJ0100-B00	WIRE BARE JUMPER 10MM	7	J021 J022 J216 J403 J615 J616 J818
41-WJ0100-B00	WIRE BARE JUMPER 10MM	6	J901 J912 J914 J916 J919A J921
41-WJ0115-B00	WIRE BARE JUMPER 11.5MM	1	J607
41-WJ0105-B00	WIRE BARE JUMPER 10.5MM	3	J214 J228 J617
41-WJ0110-B00	WIRE BARE JUMPER 11MM	5	J001 J002 J003 J012 J918
41-WJ0125-B00	WIRE BARE JUMPER 12.5MM	7	J203 J204 J205 J206 J227 J929 J930
41-WJ0130-B00	WIRE BARE JUMPER 13MM	4	J215 J220 J405 J207
41-WJ0140-B00	WIRE BARE JUMPER 14MM	1	J224
41-WJ0145-B00	WIRE BARE JUMPER 14.5MM	2	J625 J904
41-WJ0150-B00	WIRE BARE JUMPER 15MM	4	J411 J420 J421 J920
41-WJ0125-B00	WIRE BARE JUMPER 12.5MM	1	J209
41-WJ0165-B00	WIRE BARE JUMPER 16.5MM	3	J225 J611 J612
41-WJ0175-B00	WIRE BARE JUMPER 17.5MM	1	J212
41-WJ0185-B00	WIRE BARE JUMPER 18.5MM	8	J016 J017 J226 J618 J619 J620 J621 J626
41-WJ0190-B00	WIRE BARE JUMPER 19MM	4	J014 J015 J018 J908
41-WJ0195-B00	WIRE BARE JUMPER 19.5MM	1	J601
41-WJ0200-B00	WIRE BARE JUMPER 20MM	3	J013 J905 J913
45-OSC8M0-0Y0	CRYSTAL 8.0MHZ	1	X001
45-SAW395-5M0	SAW FILTER K3955M (38MHz)	1	Z101
45-TRA5M5-0Y0	CER TRAP TPS 5.5MHZ	1	X203
45-TRA6M0-0Y0	CER TRAP TPS 6.0MHZ	1	X202
45-TRA6M5-0Y0	CER TRAP TPS 6.5MHZ	1	X201
46-20598W-04X	PIN BASE *4 TJC1-4A	1	P411
46-33079W-02X	PIN BASE *2 TJC3-2A	2	P601(L) P602(R)
46-33079W-03X	PIN BASE *3 TJC3-3A	1	P002
46-33079W-04X	PIN BASE *4 TJC3-4A	1	P421
46-33079W-05X	PIN BASE *5 TJC3-5A	1	P001 P201(P1 TO P5)
46-33079W-07X	PIN BASE *7 TJC3-7A	1	P921
47-RCA023-XX1	RCA JACK 3PV YEL WHI RED/SW	2	P901 P904
62-2501MB-3UN	MAIN BD BRACKET	1	
63-B30100-BT4	S/T SCREW B 3 X 10 BT	2	MTG M.BD & FBT
63-W30100-AB4	S/T SCREW W 3 X 10 AB	8	MTG M.BD & M.BD BKT
64-P30080-104	M/C SCREW P 3 X 8	1	FOR IC301
64-P30080-104	M/C SCREW P 3 X 8	2	FOR IC602
64-P30080-104	M/C SCREW P 3 X 8	1	FOR Q411
64-P30080-104	M/C SCREW P 3 X 8	1	FOR Q414
64-P30080-104	M/C SCREW P 3 X 8	2	FOR M.BD FBT
65-Z30050-23M	NUT M 3	2	FOR IC301 FOR Q411
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	5	T461
66-382330-0B7	RIVET 2.0X3.7X3.5	2	R404
67-H27292-1A0	HEAT SINK-REGULATOR & DPC A=60	1	FOR Q414
67-H30147-0A0	HEAT SINK FOR HORIZ (2501/2909	1	FOR Q411
67-H30179-5A0	HEAT SINK	1	FOR IC301
67-H34423-4A0	HEAT SINK	1	FOR IC602
90-209770-SR1	SILICONE GREASE G-746	0.0009	
90-269080-ZU0	CLEAN COATING TC-131L 14KG/BUCKET	0.0001	
34-R829J2-0EX	COIL PL - 8.2 UH +/-5%	1	L204
47-RCA089-XX0	RCA SOCKET AV-3.2-3LK-N2	1	P903
41-WJ0090-B00	WIRE BARE JUMPER 9MM	2	L1102 L1103
41-WJ0065-B00	WIRE BARE JUMPER 6.5MM	2	D413 D414
41-WJ0150-B00	WIRE BARE JUMPER 15MM	1	L412
18-CB0563-JNX	RES. C.F. 56K OHM 1/6W +/-5%	1	R414
26-EBP102-KBX	CAP. CER 1000 PF 50V +/-10% B	1	C416
18-CB0820-JNX	RES. C.F. 82 OHM 1/6W +/-5%	2	R206A R207A
18-CB0220-JNX	RES. C.F. 22 OHM 1/6W +/-5%	2	R027 R028
26-EBP220-JCX	CAP. CER 22PF 50V +/-5% CH	2	C023 C024
54-128700-0X0	FIBRE WASHER 3.3 X 10 X 0.8	1	FOR FBT
18-CB0472-JNX	RES. C.F. 4.7K OHM 1/6W +/-5%	2	R602 R612

27-AGQ474-J0X	CAP. M.PP 0.47UF 250V +/-5%	1	C419
36-WID601-XX1	COIL CHOKE 600 UH	1	L413
41-WJ0060-B00	WIRE BARE JUMPER 6mm	2	J303 J304
26-EBP221-JCX	CAP. CER 220PF 50V +/-5% CH	1	C035
40-2970ME-MAE	P.C.B. MAIN BD	1	
18-DD0123-FN7	RES. M.F. 12K OHM 1/4W +/-1%	1	R308
36-HDR001-AX1	TRANSFOR HOR. DRIVE (NEGMATIC COIL)	1	T401
11-630MFP-0AX	TRANSISTOR IRF630MFP	1	Q414
25-BCA100-M1X	CAP. ELEC 10 UF 16V +/-20%	9	C902 C903 C905 C906 C911 C912 C917 C021 C033
25-BCA101-M1X	CAP. ELEC 100 UF 16V +/-20%	8	C219 C233 C914 C930 C930A C028 C612 C210
25-BCA102-M1X	CAP. ELEC 1000 UF 16V +/-20%	1	C245
25-BCA470-M1X	CAP. ELEC 47 UF 16V +/-20%	4	C213 C901 C904 C920
25-BCB100-M1X	CAP. ELEC 10 UF 16V +/-20%	2	C228 C235
25-BCB221-M1X	CAP. ELEC 220 UF 16V +/-20%	1	C910
25-BCB470-M1X	CAP. ELEC 47 UF 16V +/-20%	2	C229 C631
25-BCB479-M1X	CAP. ELEC 4.7 UF 16V +/-20%	9	C223 C304 C614 C615 C616 C617 C618 C619 C620
25-BDB100-M1X	CAP. ELEC 10 UF 25V +/-20%	1	C632
25-BDG102-M1X	CAP. ELEC 1000 UF 25V +/-20%	1	C623
25-BEA221-M1X	CAP. ELEC 220 UF 35V +/-20%	1	C302
25-BEA471-M1X	CAP. ELEC 470 UF 35V +/-20%	1	C301
25-BEG102-M1X	CAP. ELEC 1000UF 35V +/-20%	1	C432
25-BEG222-M1X	CAP. ELEC 2200 UF 35V +/-20%	1	C308
25-BFB101-M1X	CAP. ELEC 100 UF 50V +/-20%	3	C224 C105 C237
25-BFB109-M1X	CAP. ELEC 1 UF 50V +/-20%	2	C218 C221
25-BFB109-M1X	CAP. ELEC 1 UF 50V +/-20%	1	C307
25-BFB228-M1X	CAP. ELEC 0.22 UF 50V +/-20%	1	C216
25-BFB229-M1X	CAP. ELEC 2.2 UF 50V +/-20%	4	C601 C607 C650 C657
25-BFB478-M1X	CAP. ELEC 0.47 UF 50V +/-20%	5	C227 C246 C248 C918 C211
25-BFB479-M1X	CAP. ELEC 4.7 UF 50V +/-20%	1	C101
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C411
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C412
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C413
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C414
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C419
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR L413
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR L414
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	1	FOR Q411(C)
66-382330-0B7	RIVET 2.0X3.7X3.5	2	FOR D411
11-SC1815-YBX	TRANSISTOR 2SC1815Y	2	Q630 Q631
18-CB0223-JNX	RES. C.F. 22K OHM 1/6W +/-5%	4	R630 R631 R632 R633
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	2	R604 R611
13-A20V01-TOP	IC TCL-A20V01-TO	1	IC101
41-WJ0050-B00	WIRE BARE JUMPER 5MM	1	J230
41-WJ0075-B00	WIRE BARE JUMPER 7.5MM	1	J414
08-2918AE-NIY	ASS'Y - NICAM BD	1	
11-SC388A-MBX	TRANSISTOR 2SC388ATM (RF AMPL)	1	Q1101
13-TDA980-1AP	IC TDA9801A	1	IC1101
13-TDA987-4AP	IC TDA9874APS	1	IC1102
18-CB0100-JNX	RES. C.F. 10 OHM 1/6W +/-5%	1	R1111
18-CB0100-JNX	RES. C.F. 10 OHM 1/6W +/-5%	1	R1109
18-CB0101-JNX	RES. C.F. 100 OHM 1/6W +/-5%	2	R1114 R1115
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	1	R1106
18-CB0123-JNX	RES. C.F. 12K OHM 1/6W +/-5%	1	R1104
18-CB0223-JNX	RES. C.F. 22K OHM 1/6W +/-5%	2	R1107 R1108
18-CB0331-JNX	RES. C.F. 330 OHM 1/6W +/-5%	1	R1103
18-CB0331-JNX	RES. C.F. 330 OHM 1/6W +/-5%	1	R1112
18-CB0339-JNX	RES. C.F. 3.3 OHM 1/6W +/-5%	1	R1110
18-CB0391-JNX	RES. C.F. 390 OHM 1/6W +/-5%	1	R1105
18-CB0393-JNX	RES. C.F. 39K OHM 1/6W +/-5%	1	R1101

18-CB0563-JNX	RES. C.F. 56K OHM 1/6W +/-5%	1	R1102
18-CB0822-JNX	RES. C.F. 8.2K OHM 1/6W +/-5%	1	R1113
26-EBP102-KBX	CAP. CER 1000 PF 50V +/-10% B	4	C1101 C1107 C1122 C1124
26-EBP103-ZFX	CAP. CER 0.01UF 50V +80/-20% F	3	C1112 C1113 C1115
26-EBP104-ZFX	CAP. CER 0.1UF 50V +80%/-20%	7	C1104 C1117 C1119 C1118 C1120 C1125 C1126
26-EBP680-JZX	CAP. CER 68PF 50V +/-5% SL TUBE	1	C1108
27-MBC104-J0X	CAP. M.P.E 0.1 UF 63V +/-5%	3	C1102 C1123 C1106
34-A100K0-1IX	COIL CHOKE 10UH +/-10%	1	L1101
35-139730-00X	FERR. BEAD BF60	2	FOR L1102 & L1103
38-309820-00X	COIL I.F.T FOR VCO 76 MHz	1	T1101
41-WJ0060-B00	WIRE BARE JUMPER 6mm	5	J102 J1105 J1107 J1111 J1114
41-WJ0075-B00	WIRE BARE JUMPER 7.5MM	2	J1101 J1102
41-WJ0085-B00	WIRE BARE JUMPER 8.5MM	3	J1103 J1104 J1106
41-WJ0090-B00	WIRE BARE JUMPER 9MM	2	J103 J1110
41-WJ0100-B00	WIRE BARE JUMPER 10MM	1	J009
41-WJ0115-B00	WIRE BARE JUMPER 11.5MM	1	J1109
41-WJ0125-B00	WIRE BARE JUMPER 12.5MM	1	J1112
41-WJ0150-B00	WIRE BARE JUMPER 15MM	1	J1113
41-WJ0175-B00	WIRE BARE JUMPER 17.5MM	2	J010 J011
41-WJ0150-B00	WIRE BARE JUMPER 15MM	4	J019 J020 J101 J104
41-WJ0175-B00	WIRE BARE JUMPER 17.5MM	1	J105
45-OSC24M-5N0	CRYSTAL 24.576 MHZ	1	X1101
45-SAW926-1M0	SAW FILTER 31.5-32.0M(K9261M)	1	Z1101
46-30215H-02X	HS 1365#26 200MM TJC3-2Y/SCN-2	1	P101
46-33079W-02X	PIN BASE *2 TJC3-2A	1	P1101
25-BCB220-M1X	CAP. ELEC 22 UF 16V +/-20%	1	C1105
25-BCB470-M1X	CAP. ELEC 47 UF 16V +/-20%	1	C1111
25-BCB471-M1X	CAP. ELEC 470 UF 16V +/-20%	1	C1103
25-BFB109-M1X	CAP. ELEC 1 UF 50V +/-20%	1	C1121
08-2918AE-PAN	ASS'Y - PACKING	1	
72-2918AE-E129A	OPERATION MANUAL	1	
74-010050-40C	POLYBAG FOR POWER (10CMX50CM)	1	
74-022032-6WE	POLYBAG (22cmX32cmX0.06mm)	1	
74-130130-80HAA	POLYBAG W/SUFFOCATION WARNING	1	
75-2918LL-CC0	POLYFOAM (LL)	1	
75-2918LR-CC0	POLYFOAM (LR)	1	
75-2918UL-CC0	POLYFOAM (UL)	1	
75-2918UR-CC0	POLYFOAM (UR)	1	
76-002918-0AT	CARTON BOX	1	
08-2918AE-PWY	ASS'Y - POWER PARTS	1	
10-0RU3YX-F0X	DIODE RU3YX (FAST RECTIFIER)	1	D830
10-1N4001-EBX	DIODE 1N4001 (RECTIFIER)	1	D836
10-1N4148-ABX	DIODE 1N4148 (SWITCHING)	2	D804 D805
10-79C12V-DBX	DIODE ZENER 12V 1/2W 5%	1	D838
10-CW574C-DJX	DIODE CW574CD	1	D834
10-HER108-FBX	DIODE HER108	1	D806
10-T3SB60-H7X	DIODE ZENER T3SB60 4.0A 600V	1	DB801
11-C124ES-0BX	TRANSISTOR PDTTC124ES (NPN)	2	Q833 Q834
11-SC1815-YBX	TRANSISTOR 2SC1815Y	2	Q831 Q832
11-SK2996-0AX	TRANSISTOR 2SK2996 (MOS)	1	Q801
13-00KA78-05S	IC KA7805 5V 1A (REGULATOR)	1	IC804
13-00KA78-09S	IC KA7809 9V 1A (REGULATOR)	1	IC803
13-00TLP6-21P	PHOTO COUPLER TLP621(GRH)	1	IC802
13-44608P-40P	IC MC44608P40	1	IC801
18-BE0105-JN1	RES. C.C. 1M OHM 1/2W +/-5%	1	R805
18-CB0102-JNX	RES. C.F. 1K OHM 1/6W +/-5%	1	R809
18-CB0103-JNX	RES. C.F. 10K OHM 1/6W +/-5%	2	R844 R835A
18-CB0332-JNX	RES. C.F. 3.3k OHM 1/6W +/-5%	1	R834
18-CB0333-JNX	RES. C.F. 33K OHM 1/6W +/-5%	1	R838

18-CB0392-JNX	RES. C.F. 3.9K OHM 1/6W +/-5%	1	R807
18-CB0471-JNX	RES. C.F. 470 OHM 1/6W +/-5%	1	R807A
18-CB0472-JNX	RES. C.F. 4.7K OHM 1/6W +/-5%	1	R806
18-CD0103-JNX	RES. C.F. 10K OHM 1/4W +/-5%	1	R820
18-CD0220-JNX	RES. C.F. 22 OHM 1/4W +/-5%	1	R811A
18-CD0221-JNX	RES. C.F. 220 OHM 1/4W +/-5%	1	R833
18-CD0471-JNX	RES. C.F. 470 OHM 1/4W +/-5%	1	R811
18-CE0479-JNX	RES. C.F. 4.7 OHM 1/2W +/-5%	1	R832
18-DD0102-FNX	RES. M.F. 1K OHM 1/4W +/-1%	1	VR829
18-DD0104-FNX	RES. M.F. 0.1M OHM 1/4W +/-1%	1	R804
18-DD0392-FNX	RES. M.F. 3.9K OHM 1/4W +/-1%	1	R835
18-DD0472-FNX	RES. M.F. 4.7K OHM 1/4W +/-1%	1	R804A
18-DE0823-FNX	RES. M.F. 82K OHM 1/2W +/-1%	1	R831
18-FG0223-JHX	RES. M.O. 22K OHM 2W +/-5%	1	R839
18-FG0333-JHX	RES. M.O. 33K OHM 2W +/-5%	1	R837
18-GG0228-JHX	RES WIRE ROUND 0.22 OHM 2W 5%	1	R810
18-GJ0103-KTX	RES. CEMENT 10K OHM 5W +/-10%	1	R836
18-GJ0223-KTX	RES. CEMENT 5W 22K OHM +/-10%	1	R808
22-NTC479-XX0	NTC 4.7 OHM +/-18% NTC4.7D2-14	1	RT802
22-PTC200-XX0	POSISTOR 20 OHM (25-34)	1	RT801
25-BJG221-M1X	CAP. ELEC 220 UF 160V +/-20%	1	C835
25-BMJ221-M1X	CAP. ELEC 220 UF 400V +/-20%	1	C806
26-ABC102-KBX	CAP. CER 1000 PF 50V +/-10% B	1	C804A
26-ABC104-ZFX	CAP. CER 0.1 UF 50V +80-20% F	1	C813
26-ABC221-JZX	CAP. CER 220 PF 50V +/-5% SL	1	C841
26-AGK221-KRX	CAP. CER 220 PF 250V +/-10%	1	C830
26-AIM103-KBX	CAP. CER 0.01UF 500V +/-10% B	2	C805 C834
26-AKK221-KRX	CAP. CER 220 PF 1KV +/-10% R	1	C833
26-AMK152-KRX	CAP. CER 1500 pF 2KV +/-10% R	1	C818
26-AMK331-JZX	CAP. CER 330 PF 2KV +/-5% SL	1	C817
26-APK471-KBX	CAP. CER 470PF 400VAC +/-10% B	2	C803 C804
26-AQK472-ZFX	CAP. CER 4700PF 250VAC+80-20%F	2	C807
26-AQK472-ZFX	CAP. CER 4700PF 250VAC+80-20%F	1	C808
26-EBP101-JCX	CAP. CER 100PF 50V +/-5% CH	1	C814
26-EBP102-KBX	CAP. CER 1000 PF 50V +/-10% B	1	C844
26-EBP104-ZFX	CAP. CER 0.1UF 50V +80%/-20%	4	C842 C831 C846 C864
27-MBC104-J0X	CAP. M.P.E 0.1 UF 63V +/-5%	1	C820
27-MHW104-K0X	CAP. M.P.E 0.1 UF 400V +/-10%	1	C802A
27-RJK472-J0X	CAP. PP 4700PF 630V +/-5%	1	C809
34-R101K2-1BX	COIL CHOKE 100 UH +/-10%	2	L801 L802
35-139730-00X	FERR. BEAD BF60	3	L814 L815 L816
35-237250-00X	FERR. BEAD HF70	4	L810 L812 L811 L813
35-237250-00X	FERR. BEAD HF70	2	FOR D806
35-237250-00X	FERR. BEAD HF70	2	FOR D804
36-304090-002	LINE FILTER LCL-2826 (2501)	1	T801
36-TRF046-XX1	TRANSFORMER CONV. BCK-4201-39m	1	T803
41-WJ0065-B00	WIRE BARE JUMPER 6.5MM	1	J815
41-WJ0100-B00	WIRE BARE JUMPER 10MM	6	J805 J802G R840 L816 J804 J814
41-WJ0125-B00	WIRE BARE JUMPER 12.5MM	2	J816 J817
41-WJ0130-B00	WIRE BARE JUMPER 13MM	1	J803
41-WJ0145-B00	WIRE BARE JUMPER 14.5MM	1	J811
41-WJ0150-B00	WIRE BARE JUMPER 15MM	3	J809 J810 J812
41-WJ0175-B00	WIRE BARE JUMPER 17.5MM	2	J807 J813
41-WJ0195-B00	WIRE BARE JUMPER 19.5MM	1	J806
46-10962W-02X	PIN BASE *2 TJC2-2A	1	S803
46-28559W-02X	PIN BASE *2 TJC1-2A	1	S801
50-03150D-1GS1	FUSE 3.15AT 250VAC 5mmX20mm	1	FOR F801
64-B30100-104	M/C SCREW B 3 X 10	1	FOR Q801
64-P30080-104	M/C SCREW P 3 X 8	1	FOR IC803

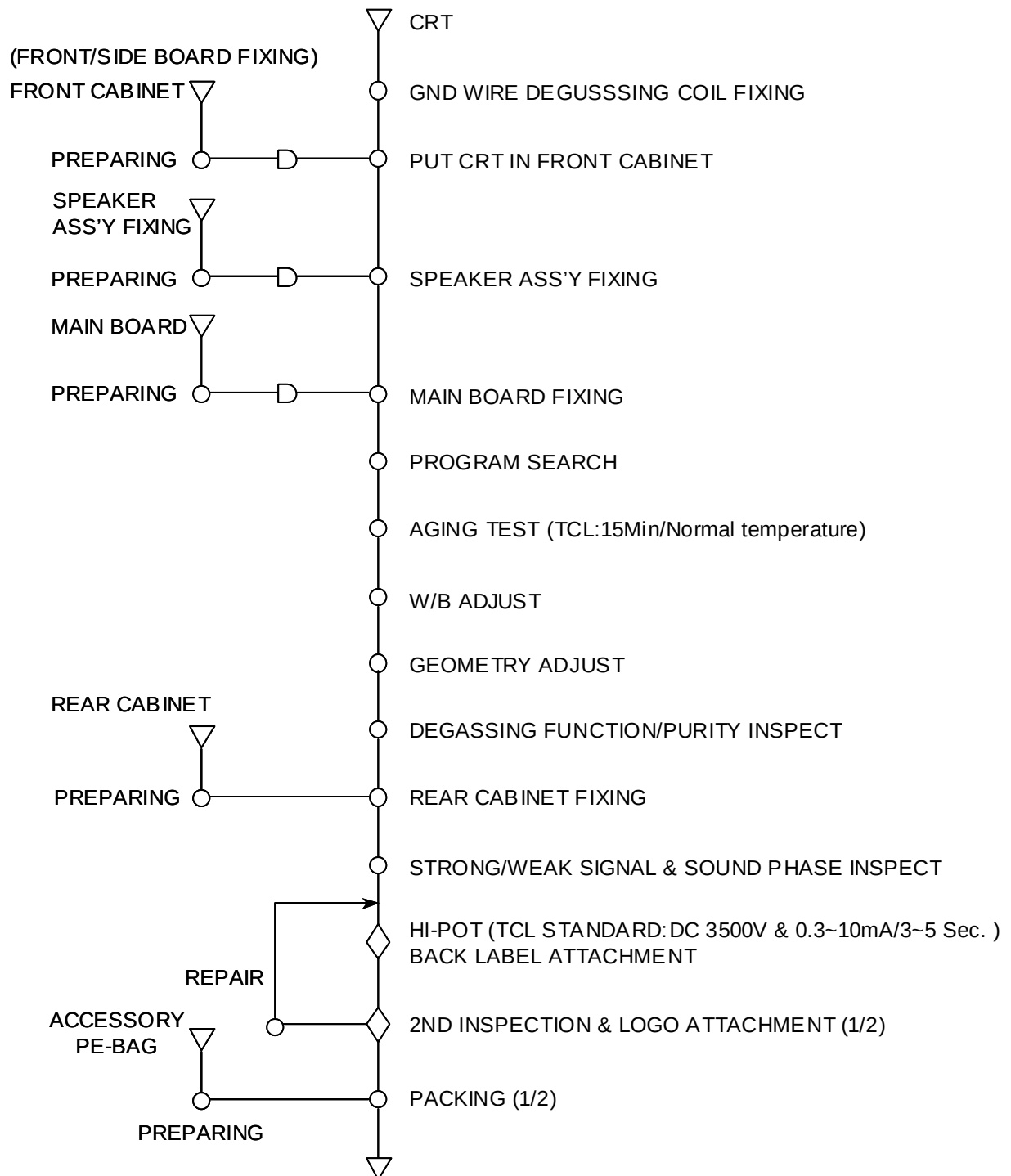
64-P30080-104	M/C SCREW P 3 X 8	1	FOR IC804
66-20516X-0B0	FUSE HOLDER	2	FOR F801
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR D830
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C835
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	4	FOR T803
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR DB801
66-382330-0B7	RIVET 2.0X3.7X3.5	2	FOR R808
66-382330-0B7	RIVET 2.0X3.7X3.5	2	FOR R836
67-H27292-3A0	HEAT SINK	1	FOR IC803
67-H27292-3A0	HEAT SINK	1	FOR IC804
71-DYP000-TZ1	HEAT SINK LABEL	1	STICK ON HEAT SINK
90-209770-SR1	SILICONE GREASE G-746	0.0006	
67-H35984-2A0	HEAT SINK	1	FOR Q801
11-SC2688-LAX	TRANSISTOR 2SC2688L (NPN)	1	Q830
10-0FR104-FBX	DIODE FR104	2	D802 D833
10-HS5V6B-DBX	DIODE 500mW 5.6HSB	1	D840
10-79C8V2-DBX	DIODE ZENER 8V2 1/2W 5%	1	D841
18-KF0825-JH3	RES. H.VOLT.CC 8.2M OHM 1W +/-5%	1	R812A
26-APK222-ME4	CAP. CER 2200PF 400VAC+/-20% E	1	C816
27-AQT224-MVH	CAP. M.PP 0.22 UF 250VAC 20%	2	C801 C802
25-BCB470-MIX	CAP. ELEC 47 UF 16V +/-20%	3	C849 C845 C863
25-BCB471-MIX	CAP. ELEC 470 UF 16V +/-20%	1	C843
25-BCB479-MIX	CAP. ELEC 4.7 UF 16V +/-20%	1	C812
25-BEA102-MIX	CAP. ELEC 1000 UF 35V +/-20%	1	C832
25-BEB101-MIX	CAP. ELEC 100 UF 35V +/-20%	1	C840
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	4	FOR T801
41-WJ0085-B00	WIRE BARE JUMPER 8.5MM	1	J802F
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C801
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	2	FOR C802
66-343730-0B0	HOLLOW RIVET 1.6X3.0XL3.2	1	FOR C832
10-0RU3AM-F0X	DIODE RU3AM (FAST RECOVERY)	1	D831
08-2918AE-RCN	ASS'Y - REAR CABINET	1	
54-114000-00X	FELT TAPE (150mmX19mmX0.3mm)	10	STICK ON R.CAB
55-2918RC-0CNAA	REAR CABINET	1	
58-2918MP-0UI9D	PLATE MODEL NO.	1	
58-2918RI-6UI9A	INLAY REAR AV	1	
59-130460-00X	RUBBER PAD (25mmX7mm)	2	STICK ON R.CAB(FOOTING)
63-B40250-AB2	S/T SCREW B 4 X 25 AB	4	MTG FRONT & R.CAB
63-B40250-AB2	S/T SCREW B 4 X 25 AB	6	MTG FRONT & R.CAB
63-F30100-BT3	S/T SCREW F 3 X 10 BT (BLACK)	2	MTG RCA JACK & R.CAB
08-2918AE-SIY	ASS'Y - SIDE AV BD	1	
40-2518SM-SIB	P.C.B. SIDE AV BD	1	
41-WJ0050-B00	WIRE BARE JUMPER 5MM	4	C961 C951 C954 C955
41-WJ0050-B00	WIRE BARE JUMPER 5MM	1	Q962(B) TO (E)
41-WJ0050-B00	WIRE BARE JUMPER 5MM	1	Q963(B) TO (E)
41-WJ0060-B00	WIRE BARE JUMPER 6mm	1	C962
41-WJ0060-B00	WIRE BARE JUMPER 6mm	1	R952
41-WJ0060-B00	WIRE BARE JUMPER 6mm	1	R953
41-WJ0060-B00	WIRE BARE JUMPER 6mm	1	J956
41-WJ0060-B00	WIRE BARE JUMPER 6mm	1	IC951(2) TO (4)
41-WJ0080-B00	WIRE BARE JUMPER 8 MM	1	IC951 (8) TO (11)
41-WJ0080-B00	WIRE BARE JUMPER 8 MM	1	R964
46-30951H-07X	HS UL 2468 #24 420 3C/TS 4C/TS	1	P902
47-RCA041-XX0	JACK RCA 3PH H=8MM YL WH RD/SW	1	P951
47-SVI002-XX0	Y/C SOCKET VERTICAL TYPE	1	P961
02-2918AE-RMY	ASS'Y - REMOTE HANDSET	1	

PART VIII. Working Guide

DESIGN
LEXUS SUN

WORKING GUIDE

VERSION
1.0



PART IX. Technical Support List

Document No.: _____

Date: _____

Technical Support List

➡ Model No. : _____ ➡ Sale Destination: _____

➡ Shipment format: ☐ SKD ☐ CKD ☐ CBU

➡ Product Safety Requirement: _____ ➡ ODF No.: _____

<u>ITEM</u>	<u>OPTION</u>	<u>REMARK</u>
1. BOM	【Yes ☐ No ☐ 】	
2. Working Guide	【Yes ☐ No ☐ 】	
3. Product Specification	【Yes ☐ No ☐ 】	
4. Service manual	【Yes ☐ No ☐ 】	
5. Engineering sample	1 PCS (Dismantle) 【Yes ☐ No ☐ 】	
6. Cosmetic and Quality	1PCS (Not Dismantle) 【Yes ☐ No ☐ 】	
7. Circuit Diagram	【Yes ☐ No ☐ 】	
8. Special Equipment/JIGS	【Yes ☐ No ☐ 】	
9. Factory Remote Control Hand Set	【Yes ☐ No ☐ 】	QTY: PCS
10. Alignment Procedure	【Yes ☐ No ☐ 】	
11. Service training	【Yes ☐ No ☐ 】	
12. Key Component (CRT 、 FBT、 IC) Specification	【Yes ☐ No ☐ 】	

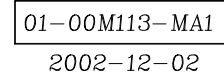
Prepared By: _____ (Marketing Dept.)

Confirmed By: _____ (Overseas Manufacturing Management Dept.)

TCL Overseas Manufacturing Management Dept.

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Nicam Option

SERVICE MANUAL M113 (SUPER ONE CHIP)

TCL



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